

pb FIND THE SUCCESSIVE APPROXIMATION ADC OUTPUT FOR A 4 BIT CONVERTER TO A 3.217 V INPUT IF THE REFERENCE IS 5V.

INPUT = 3.217 V
REF = 5V

SET $\rightarrow N=1$

I

$$V_f = V_{ref} \times 2^{-n} = 5 \times 2^{-1} = \frac{5}{2} = 2.5V$$

$V_{in} = 3.217V$, $V_f = 2.5V$

$V_{in} > V_f \longrightarrow b_1 = 1$

II

INPUT = 3.217V, SET $n=2$

REF = 5V

$$V_f = \left(V_{ref} \times 2^{-n} + \frac{V_{ref}}{2} \right) = 5 \times 2^{-2} + \frac{5}{2} = 3.75V$$

COMPARE $V_{in} 3.217V$ WITH $V_f 3.75V$

$V_f > V_{in} \longrightarrow b_2 = 0$

III

INPUT =

REF =

compare

IV

INPUT

REF =

III INPUT = 3.217 V SET $n=3$

$$V_{ref} = 5V \rightarrow V_f = V_{ref} \times 2^{-n} + \frac{V_{ref}}{2}$$
$$= 5 \times 2^{-3} + \frac{5}{2} = 3.125V$$

COMPARE V_{IN} 3.217 V WITH V_f 3.125 V

$$V_{IN} > V_f \rightarrow b_3 = 1$$

b_1	b_2	b_3	b_4
1	0	1	0

2

IV INPUT = 3.217 V SET $n=4$

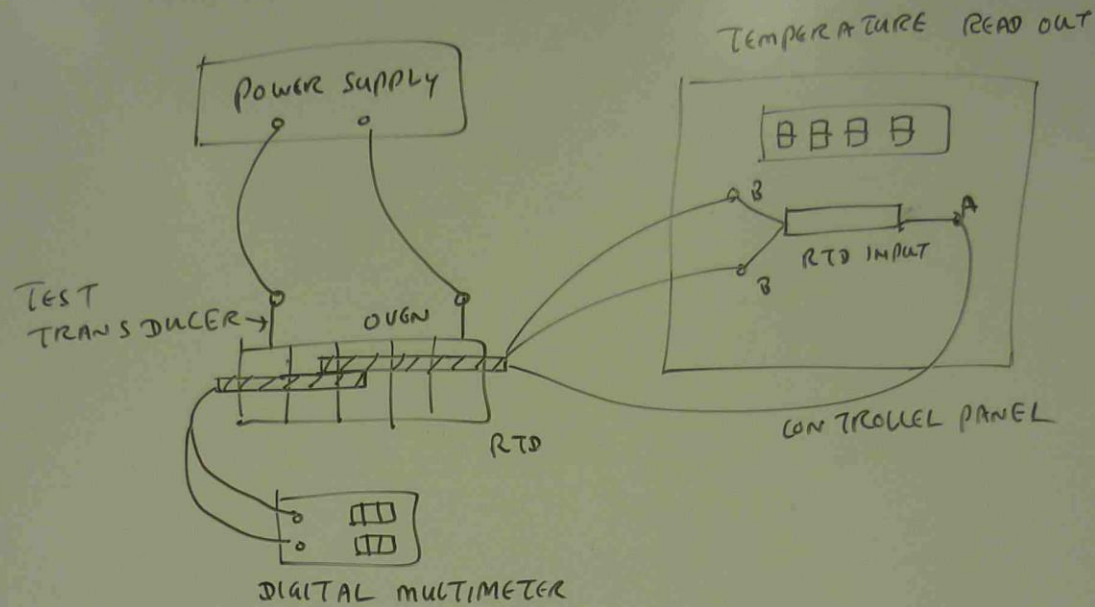
$$V_{ref} = 5V$$
$$V_f = V_{ref} \times 2^{-n} + \frac{V_{ref}}{2}$$

$$= 5 \times 2^{-4} + \frac{5}{2} = 3.4375V$$

COMPARE V_{IN} 3.217 V WITH V_f 3.4375 V

$$V_f > V_{IN} \rightarrow b_4 = 0$$

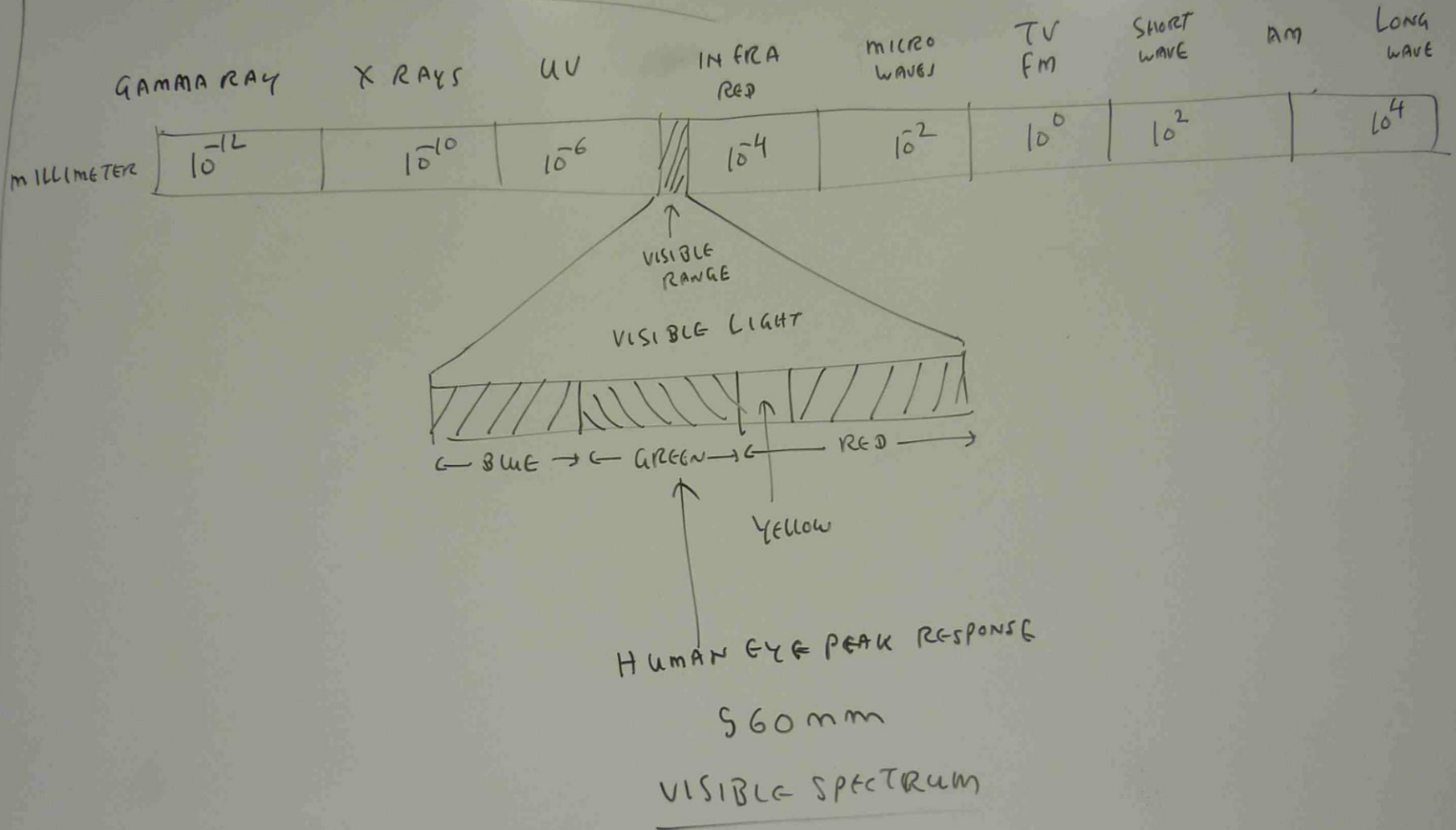
TEMPERATURE MEASUREMENT



Equipment

- TEMPERATURE TRANSDUCER (RTD, THERMISTOR, THERMOCOUPLE)
- DIGITAL MULTIMETER
- HEATING OVEN
- UNIVERSAL CONNECTING BLOCK
- DC POWER SUPPLY - 2A
- TEMPERATURE CONTROL PANEL
- CONNECTION LEAD - 4mm BANANA LEAD

LIGHT AND LIGHT TRANSDUCERS



AM

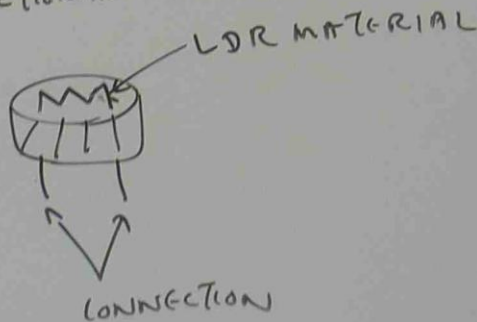
LONG
WAVE

LO⁴

LDR (LIGHT DEPENDENT RESISTOR) → PHOTO CONDUCTIVE CELL
(0.6φ × 15mm)

OPERATION

LIGHT FALLING ON THE SURFACE OF LDR
CAUSES A DECREASE IN RESISTANCE THAT IS
PROPORTIONAL TO THE AMOUNT OF LIGHT



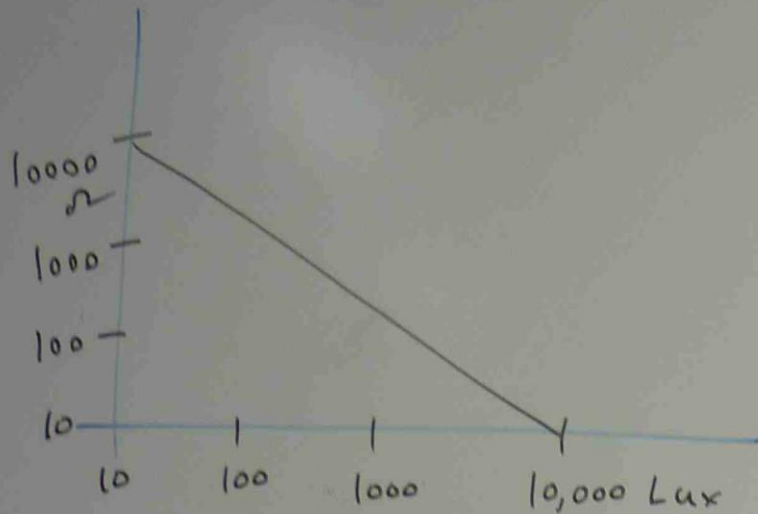
DARK VALUE $R_D = \text{minimum } 10\text{M}\Omega$ (TOTAL DARKNESS)

LIGHT VALUE $R_L = 75 - 300\Omega$ (MEASURED AT 1000 LUX)

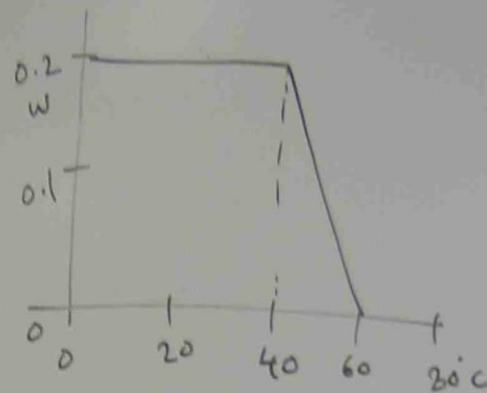
PERMISSIBLE VOLTAGE = $E_{\text{MAX}} = 150\text{V}$

AMBIENT TEMPERATURE = $-20 \rightarrow +60^\circ\text{C}$

CAPACITANCE
6 pF



RESISTANCE VALUE AS FUNCTION
OF LIGHT INTENSITY



DERATING CURVE

TWO TYPES OF LDRs ARE AVAILABLE AND
THEIR RESPONSES ARE AS FOLLOWS

CADMIUM	SELENIDE	$C_{dse} = 735 \text{ nm}$ (INFRARED)
CADMIUM	SULPHIDE	$C_{ds} = 530 \text{ nm TO } 600 \text{ nm}$ (EYE)

DIGITAL TO ANALOG CONVERTERS (DACs)

$$V_{out} = V_R \left[b_1 \bar{2}^{-1} + b_2 \bar{2}^{-2} + \dots + b_m \bar{2}^{-m} \right]$$

V_{out} = ANALOG OUTPUT VOLTAGE

V_R = REFERENCE VOLTAGE

b_1, b_2, \dots, b_m = m BIT BINARY WORD

$$V_{out} = \frac{N}{2^m} V_{ref}$$

Pb WHAT IS THE OUTPUT VOLTAGE OF A 10 BIT DAC WITH A 10V REFERENCE IF THE INPUT IS (a) 0010110101
(b) 20FH

WHAT INPUT IS NEEDED FOR 6.5V OUTPUT

$$\begin{aligned}
 (a) \quad V_{out} &= V_{REF} \left(b_1 \bar{2}^1 + b_2 \bar{2}^2 + \dots + b_n \bar{2}^n \right) \\
 &= 10 \left(0 \times \bar{2}^1 + 0 \times \bar{2}^2 + 1 \times \bar{2}^3 + 0 \times \bar{2}^4 + 1 \times \bar{2}^5 + 1 \times \bar{2}^6 + 0 \times \bar{2}^7 + 1 \times \bar{2}^8 + 0 \times \bar{2}^9 + 1 \times \bar{2}^{10} \right) \\
 &= 10 \times 0.1767 \\
 &= 1.767 \text{ VOLT}
 \end{aligned}$$

$$\begin{aligned}
 (b) \quad 20 \text{ fH} &= \underset{\substack{\uparrow \uparrow \uparrow \\ 2 \quad 1 \quad 0}}{2 \times 16^2 + 0 \times 16^1 + 19 \times 16^0} = \frac{527}{10}
 \end{aligned}$$

$$V_{out} = \frac{N}{2^n} V_{ref} = \frac{527}{2^{10}} \times 10 = 5.14648 \text{ V}$$

$$V_{out} = \frac{N}{2^n} V_{ref}$$

$$G.S = \frac{N}{2^{10}} \times 10$$

$$N = \frac{G.S \times 2^{10}}{10} = 665.6 \text{ V}$$

$$\Delta V_{out} = V_r \times 2^{-n}$$

ΔV_{out} = SMALLEST OUT PUT CHANGE

V_r = REFERENCE VOLTAGE

n = NUMBER OF BITS IN THE WORD.

pb 5 BIT D/A CONVERTER, 10V REFERENCE

$$\Delta V_{out} = V_r \times 2^{-n}$$

$$\Delta V_{out} = 10 \times 2^{-5} = 0.3125 \text{ V/BIT}$$

Ph FIND THE SUCCESSIVE APPROXIMATION ADC OUTPUT FOR
A 4 BIT CONVERTER TO A 3.217 V INPUT IF
THE REFERENCE IS 5V.

$$(V_{in}) \quad I/P = 3.217V$$

$$REF = 5V$$

SET $n=1$

$$V_f = V_{REF} \times \frac{1}{2} = 5 \times \frac{1}{2} = \frac{5}{2} = 2.5V$$

V_{in} COMPARES WITH V_f

$$\begin{array}{ccc} V_{in} & > & V_f \\ 3.217 & & 2.5 \end{array} \longrightarrow b_1 = 1$$

II SET $n=2$

$$V_f = V_{REF} \times \frac{1}{2} + \frac{V_{REF}}{2}$$

$$= 5 \times \frac{1}{2} + \frac{5}{2}$$

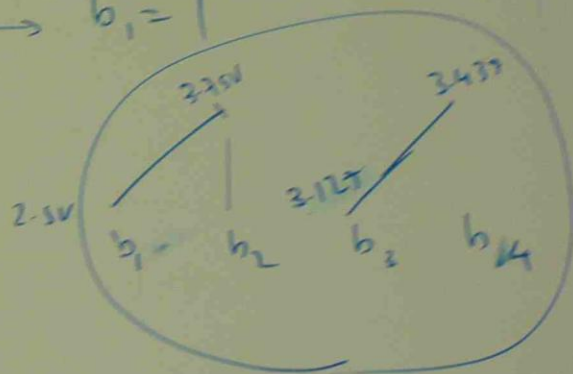
$$= 3.75V$$

V_{in} COMPARES WITH V_f

$$3.217$$

$$3.75$$

$$V_{in} < V_f \rightarrow b_2 = 0$$



III SET $m=3$

$$V_f = V_{ref} \times 2^{-m} + \frac{V_{ref}}{2}$$

$$= 5 \times 2^{-3} + \frac{5}{2}$$

$$= 3.127 \text{ V}$$

V_{IN} COMPARES WITH V_f
3.217 3.127

$$V_{IN} > V_f \rightarrow b_3 = 1$$

IV SET $m=4$

$$V_f = V_{REF} \times 2^{-m} + \frac{V_{REF}}{2}$$

$$= 5 \times 2^{-4} + \frac{5}{2}$$

$$= 3.437 \text{ V}$$

V_{IN} COMPARES WITH V_f

3.217

3.437

$$V_{IN} < V_f \rightarrow b_4 = 0$$

$b_1 \ b_2 \ b_3 \ b_4$

1 0 1 0

RAMP A/D (RAMP ANALOG TO DIGITAL CONVERTER)

THE RAMP TYPE A/D CONVERTERS ESSENTIALLY COMPARE THE INPUT VOLTAGE AGAINST A LINEARLY INCREASING RAMP VOLTAGE.

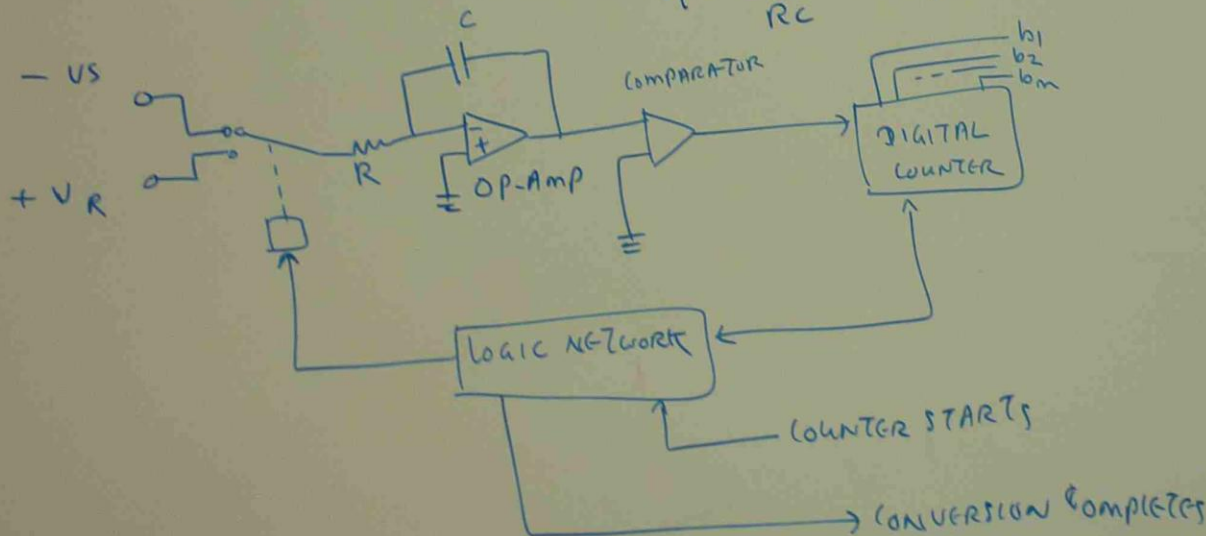
DUAL SLOPE A/D

THE INPUT SIGNAL TO DRIVE THE INTEGRATOR FOR A FIXED TIME (T_1)

GENERATING OUTPUT

$$V_1 = \frac{1}{RC} \int V_x dt$$

$$V_1 = \frac{1}{RC} T_1 V_x$$



A DUAL SLOPE A/D C WAS
AN OP-AMP INTEGRATOR,
COMPARATOR AND ASSOCIATED
DIGITAL CIRCUITS.

DIGITAL TO ANALOG CONVERTERS (DACs)

$$V_{out} = V_R [b_1 2^{-1} + b_2 2^{-2} + \dots + b_m 2^{-m}]$$

V_{out} = ANALOG OUTPUT VOLTAGE

V_R = REFERENCE VOLTAGE

b_1, b_2, \dots, b_m = m BIT BINARY WORD

$$V_{out} = \frac{N}{2^m} V_{ref}$$

pb WHAT IS THE OUTPUT VOLTAGE OF A 10 BIT DAC WITH A 10V REFERENCE IF THE INPUT IS (a) 001010101

(b) 20FH

(c) WHAT INPUT IS NEEDED FOR 6.5V OUTPUT

$$\begin{aligned} \text{(a)} \quad V_{out} &= V_{REF} [b_1 2^{-1} + b_2 2^{-2} + \dots + b_m 2^{-m}] \\ &= 10 [0 \times 2^{-1} + 0 \times 2^{-2} + 1 \times 2^{-3} + 0 \times 2^{-4} + 1 \times 2^{-5} + 1 \times 2^{-6} + 0 \times 2^{-7} + 1 \times 2^{-8} + 0 \times 2^{-9} + 1 \times 2^{-10}] \\ &= 10 \times 0.1767 \\ &= 1.767 \text{ VOLT} \end{aligned}$$

$$\begin{aligned} \text{(b)} \quad 20FH &= 2 \times 16^2 + 0 \times 16^1 + 15 \times 16^0 = 527_{10} \\ \uparrow \uparrow \uparrow \\ 2 \quad 1 \quad 0 \end{aligned}$$

$$V_{out} = \frac{N}{2^m} V_{ref} = \frac{527}{2^{10}} \times 10 = 5.14648 \text{ V}$$

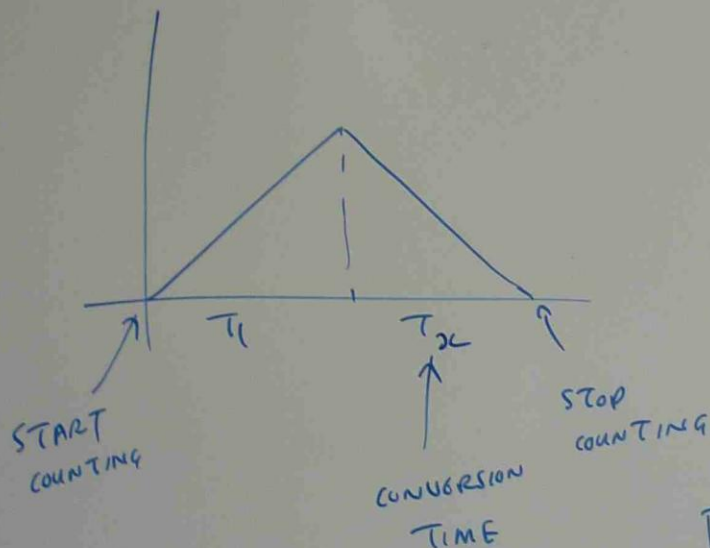
$$\text{(c)} \quad V_{out} = \frac{N}{2^m} V_{ref}$$

$$6.5 = \frac{N}{2^{10}} \times 10$$

$$N = \frac{6.5 \times 2^{10}}{10} = 665.6 \approx 665_{10}$$

pb

A DUAL SLOPE ADC AS SHOWN IN FIGURE HAS $R = 100 \text{ k}\Omega$, $C = 0.01 \mu\text{F}$.
THE REFERENCE IS 10 V AND THE FIXED INTEGRATION TIME IS 10 ms .
FIND THE CONVERSION TIME FOR 6.8 V INPUT.



$$V_1 = \frac{1}{R_C} T_1 V_R$$

$$V_1 = \frac{1}{100 \times 10^3 \times 0.01 \times 10^{-6}} \times 10 \times 10^{-3} \times 6.8$$

$$V_1 = 6.8 \text{ V}$$

$$T_{\text{CONVERSION}} = \frac{T_1 \times V_1}{V_R}$$

$$= \frac{10 \times 10^{-3} \times 6.8}{10}$$

$$= 6.8 \text{ ms}$$

GENERAL CHARACTERISTICS

(1) INPUT Common LEVEL $0 \rightarrow 10$, $0 \rightarrow 5\text{ V}$, $-10 \rightarrow +10\text{ V}$

(2) OUTPUT

A PARALLEL (OR) SERIAL BINARY WORD THAT IS ENCODING OF THE ANALOG INPUT

(3) REFERENCE

A STABLE, LOW RIPPLE SOURCE AGAINST WHICH THE CONVERSION IS PERFORMED

(4) POWER SUPPLY

GENERALLY A BIPOLAR $\pm 12\text{ V}$ TO $\pm 15\text{ V}$ SUPPLY IS REQUIRED FOR THE ANALOG AMPLIFIERS AND COMPARATORS AND A $+5\text{ V}$ SUPPLY FOR DIGITAL CIRCUITRY

(5) DIGITAL SIGNAL

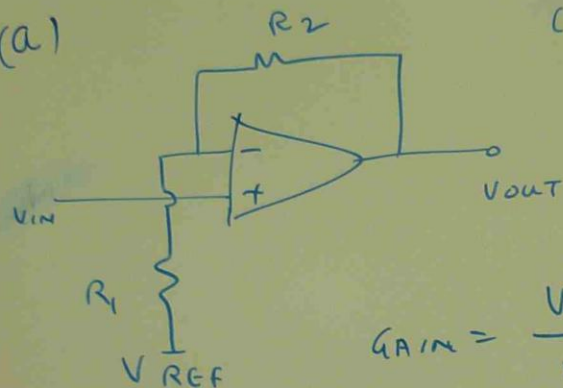
THE ADC USUALLY PROVIDES A HIGH LEVEL ON ANOTHER LINE AS AN INDICATOR.

PB

A MEASUREMENT OF TEMPERATURE USING A SENSOR THAT OUTPUTS $6.5 \text{ mV}/^\circ\text{C}$ MUST MEASURE TO 100°C . A 6 BIT ADC WITH A 10 V REFERENCE IS USED (a) DEVELOP AC CIRCUIT TO INTER FACE WITH THE SENSOR AND ADC

(b) FIND THE TEMPERATURE RESOLUTION.

(a)



$$\text{GAIN} = \frac{R_F}{R_I}$$

$$\text{GAIN} = \frac{V_{\text{OUT}}}{V_{\text{IN}}}$$

$$\begin{aligned} V_{\text{IN}} &= V/^\circ\text{C} \times \text{TEMPERATURE MEASURE} \\ &= 6.5 \times 10^{-3} \times 100 \\ &= 0.65 \text{ V} \end{aligned}$$

$$V_{\text{OUT}} = V_R (a_1^{-1} + a_2^{-2} + a_3^{-3} + \dots - a_6^{-6})$$

$$a_1 = 1, a_2 = 1, a_3 = 1, \dots, a_6 = 1$$

$$\begin{aligned} V_{\text{OUT}} &= V_R (1 \times 2^{-1} + 1 \times 2^{-2} + \dots + 1 \times 2^{-6}) \\ &= 10 \left(\frac{1}{2} + \frac{1}{2^2} + \frac{1}{2^3} + \frac{1}{2^4} + \frac{1}{2^5} + \frac{1}{2^6} \right) \\ &= 9.84375 \text{ V} \end{aligned}$$

$$GAIN = \frac{V_{out}}{V_{in}} = \frac{9.84375}{0.65} = 15.14 \quad (b)$$

$$GAIN = \frac{R_F}{R_1}$$

$$R_F = R_1 + R_2$$

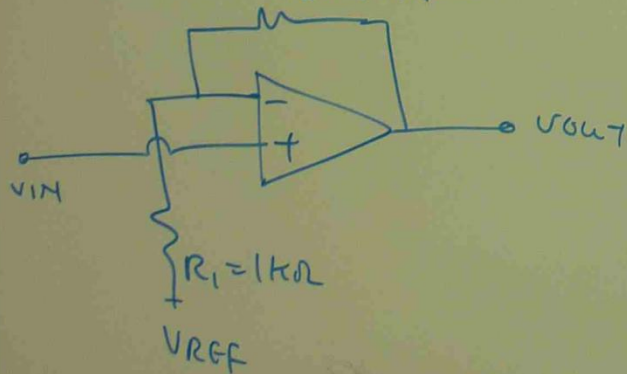
$$15.14 = R_1 + R_2$$

$$\text{SET } R_1 = 1k\Omega$$

$$15.14 = 1 + R_2$$

$$R_2 = 14.14k\Omega$$

$$R_2 = 14.14k\Omega$$



TEMPERATURE RESOLUTION

$$\Delta T = \frac{\Delta V_T}{V/^\circ C}$$

$$V/^\circ C = 6.5 \text{ mV}/^\circ C$$

$$\Delta V_T = ?$$

$$\Delta V_T = \frac{\Delta V}{GAIN}$$

$$\Delta V = V_R \times 2^{-n}$$

$$\Delta V = V_R \times 2^{-n}$$

$$\Delta V = 10 \times 2^{-6}$$

$$= 0.15625V$$

15.14 (b)

TEMPERATURE RESOLUTION

$$\Delta T = \frac{\Delta V_T}{V/^\circ\text{C}}$$

$$V/^\circ\text{C} = 6.5 \text{ mV}/^\circ\text{C}$$

$$\Delta V_T = ?$$

$$\Delta V_T = \frac{\Delta V}{\text{GAIN}}$$

$$\Delta V = V_R \times 2^{-n}$$

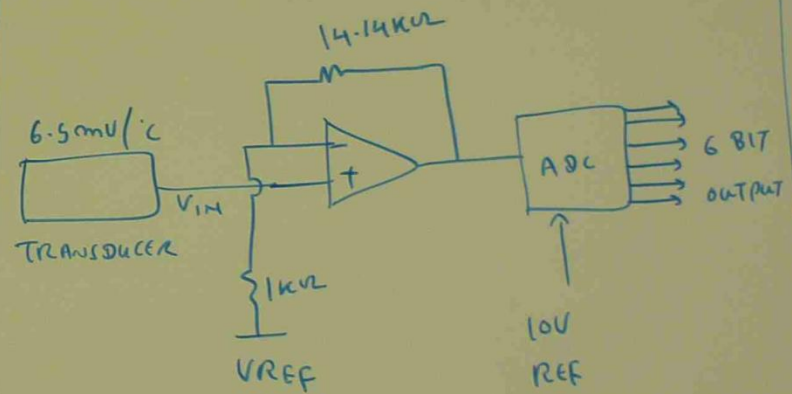
$$\Delta V = V_R \times 2^{-n}$$

$$\Delta V = 10 \times 2^{-6}$$

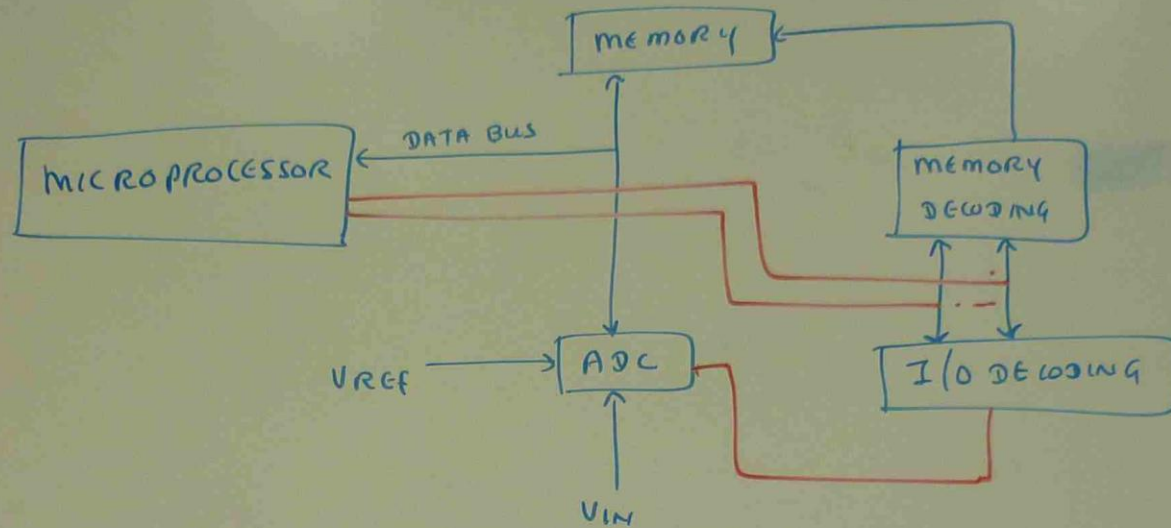
$$= 0.15625 \text{ V}$$

$$\Delta V_T = \frac{\Delta V}{\text{GAIN}} = \frac{0.15625}{15.14} = 0.01012 \text{ V}$$

$$\Delta T = \frac{\Delta V_T}{V/^\circ\text{C}} = \frac{0.01012}{6.5 \times 10^{-3}} = 1.59^\circ\text{C}$$



MICROPROCESSOR COMPATIBLE ADC



A whole line of ADC have been developed that interface easily with microprocessor based computers.

The ADCs have built in tristate outputs so that they can be connected directly to data bus of computer.

Data from ADC is only placed on data bus lines when

the computer issues an appropriate enable command (READ).

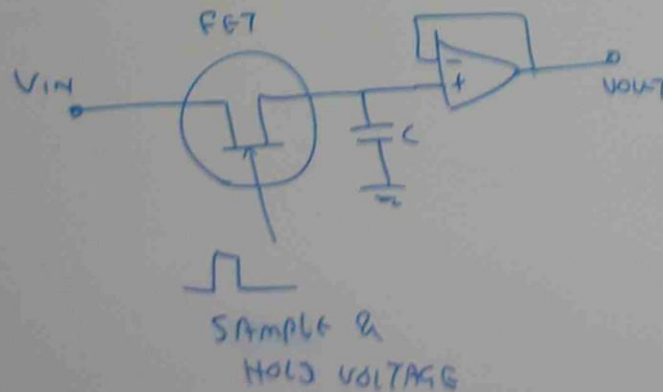
DATA INPUT BOARD

DATA INPUT BOARD, ADDRESSING, DECODING, BUS INTERFACE
REFERENCE FOR CONVERTERS.

SAMPLE AND HOLD

THE ADC REQUIRES A CERTAIN LENGTH OF TIME TO
DETERMINE THE APPROPRIATE DIGITAL OUTPUT FROM AN
ANALOG INPUT (TIME A FEW $\mu s \rightarrow ms$)

DURING CONVERSION — IT IS NECESSARY TO FREEZE
THE INPUT SIGNAL. THIS CAN BE
DONE USING A SPECIAL OP-AMP
CIRCUIT CALLED SAMPLE & HOLD



WHEN THE FET IS
DRIVEN TO THE
'OPEN' STATE,
THE VOLTAGE ON THE
CAPACITOR WILL HOLD
THE LAST VALUE
BEFORE OPEN STATE OF
FET OCCURS.

DATA ACQUISITION SYSTEM

COMPUTER CONTROLS MANY VARIABLES.

COMPUTER PERIODICALLY SAMPLES THE VALUE OF A VARIABLE.

EVALUATE IT ACCORDING TO PROGRAMMED CONTROL

OPERATIONS AND OUTPUTS AN APPROPRIATE CONTROLLING

SIGNAL TO THE FINAL CONTROL ELEMENT

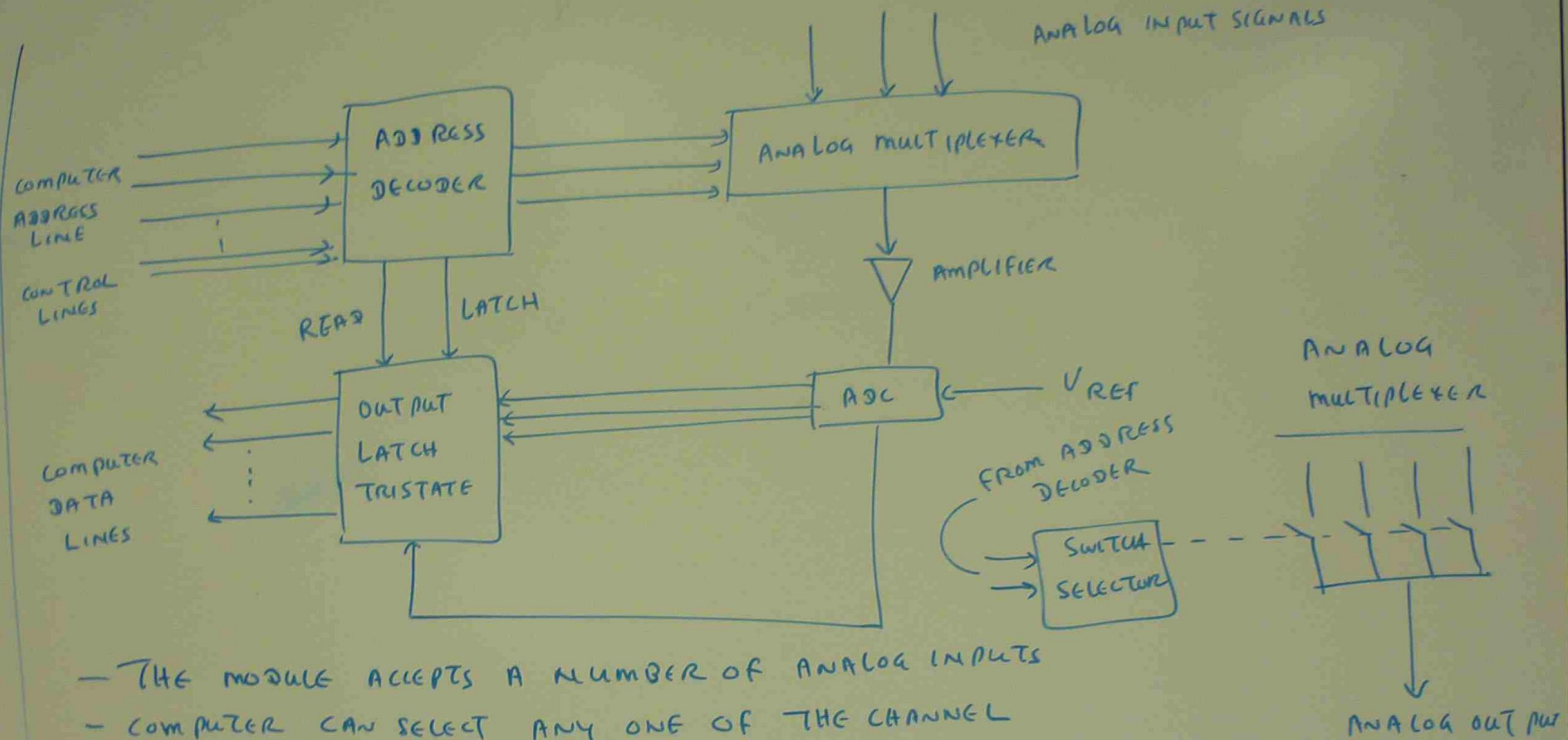
(SAMPLE | EVALUATE | OUTPUT)

INTERFACE

NECESSARY AMPLIFIERS & WRITE THE PROGRAMS

REQUIRED TO PUT TOGETHER AN INTERFACE

TO SOME COMPUTERS FOR A PROCESS APPLICATION.



- THE MODULE ACCEPTS A NUMBER OF ANALOG INPUTS
- COMPUTER CAN SELECT ANY ONE OF THE CHANNEL UNDER PROGRAM CONTROL FOR INPUT DATA
- DAS (DATA ACQUISITION SYSTEM) ACCEPTS THE INPUT FROM THE COMPUTER VIA ADDRESS LINES.

AMPLIFIER

VARIABLE GAIN AMPLIFIER

ADC

THE ADC CONVERTER ACCEPTS THE VOLTAGES THAT SPAN A SPECIFIC RANGE AS PROVIDED BY THE PRECEEDING SIGNAL CONDITIONING

SAMPLE & HOLD

HOLD THE VALUE DURING THE CONVERSION.

HARDWARE PROGRAMMING

ADDRESS SELECTION, AMPLIFIER GAIN, DIFFERENTIAL SINGLE - ENDED OPERATION.

SELECT BY JUMPERS / PINS.

SOFTWARE PROGRAMMING

USE DATA MODULES.