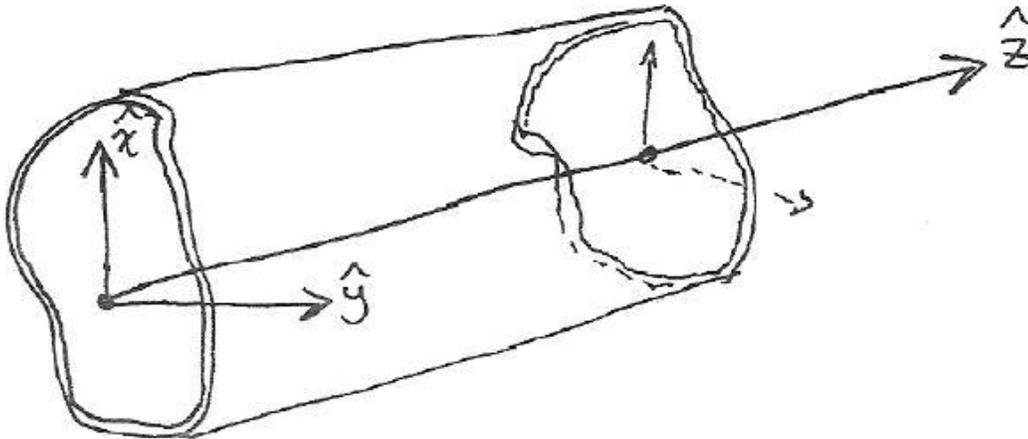


## LECTURE NOTES 10

### WAVE GUIDES and GUIDED EM WAVES

We consider/investigate the conditions under which *EM* waves can propagate when confined to the interior of some kind of “hollow” pipe – also known as a wave guide. In the real world wave guides consisting of *e.g.* rectangular, cylindrical, or arbitrarily cross-section shaped conducting and/or superconducting hollow metal pipes can be used to transport *EM* waves and *EM* energy in the radio and microwave region of the *EM* spectrum, whereas, *e.g.* glass or plastic optical fibers act as wave guides in the infrared, visible and even the UV portions of the *EM* spectrum.

We consider first the simplest type of a wave guide – a perfect conductor ( $\sigma_c = \infty$ ,  $\rho_c = 1/\sigma_c = 0$ ) such that inside the walls of the perfect conductor:  $\vec{\tilde{E}} = 0$  &  $\vec{\tilde{B}} = 0$ .  
*n.b.* in a perfect conductor  $\vec{\tilde{E}}(\vec{r}, t) = 0$  and by Faraday’s Law, if  $\vec{\nabla} \times \vec{\tilde{E}}(\vec{r}, t) = 0 \Rightarrow \partial \vec{\tilde{B}}(\vec{r}, t) / \partial t = 0$ .  
 So if  $\vec{\tilde{B}}(\vec{r}, t = 0) = 0$ , it will remain  $= 0 \forall t$ . A superconductor is a perfect conductor with  $\vec{\tilde{B}}(\vec{r}, t) = 0$  inside it (magnetic flux is expelled from a SC material – known as the Meissner effect).



The boundary conditions at the inner walls of a perfect conductor are:

$$\boxed{\oint_c \vec{\tilde{E}} \cdot d\vec{\ell} = -\frac{d}{dt} \int_s \vec{\tilde{B}} \cdot d\vec{a} = 0} \Rightarrow (1) \text{ Tangential } \vec{\tilde{E}} \text{ continuous: } \vec{\tilde{E}}_{\parallel} = 0 \text{ (since } \vec{\tilde{E}}_{\parallel}^{\text{inside}} = 0)$$

$$\boxed{\oint_s \vec{\tilde{B}} \cdot d\vec{a} = 0} \Rightarrow (2) \text{ Normal } \vec{\tilde{B}} \text{ continuous: } \vec{\tilde{B}}_{\perp} = 0 \text{ (since } \vec{\tilde{B}}_{\perp}^{\text{inside}} = 0)$$

Note that free surface charges  $\sigma_{free}$  and free surface currents  $\vec{K}_{free}$  will be induced on the inner surfaces of this perfectly-conducting wave guide so as to “enforce” these boundary conditions.

We are interested in/seek monochromatic/single-frequency plane traveling wave-type solutions - that propagate down the inside of the wave guide, *e.g.* in the  $+\hat{z}$  direction of the above figure. Generically, these must be of the form:

$$\begin{aligned}\tilde{\vec{E}}(x, y, z, t) &= \tilde{\vec{E}}_o(x, y) e^{i(k_z z - \omega t)} \\ \tilde{\vec{B}}(x, y, z, t) &= \tilde{\vec{B}}_o(x, y) e^{i(k_z z - \omega t)}\end{aligned}$$

**n.b.** for the cases of interest to us, the wave number  $k_z$  will turn out to be real.

In the interior region of the wave guide, away from (*i.e.* not inside) the walls, Maxwell's equations must be satisfied, which, for empty space or *e.g.* air with  $\epsilon_{air} \approx \epsilon_o$  and  $\mu_{air} \approx \mu_o$  are:

$$\begin{aligned}(1) \text{ Gauss' Law: } & \vec{\nabla} \cdot \tilde{\vec{E}} = 0 & (2) \text{ No magnetic charges/monopoles: } & \vec{\nabla} \cdot \tilde{\vec{B}} = 0 \\ (3) \text{ Faraday's Law: } & \vec{\nabla} \times \tilde{\vec{E}} = -\partial \tilde{\vec{B}} / \partial t & (4) \text{ Ampere's Law: } & \vec{\nabla} \times \tilde{\vec{B}} = \epsilon_o \mu_o \partial \tilde{\vec{E}} / \partial t = (1/c^2) \partial \tilde{\vec{E}} / \partial t\end{aligned}$$

The question then is, what restrictions arising from the boundary conditions (1)  $\tilde{E}^{\parallel} = 0$  and (2)  $\tilde{B}^{\perp} = 0$  are imposed on  $\tilde{\vec{E}}$  and  $\tilde{\vec{B}}$  in satisfying Maxwell's equations (1) – (4) above?

Note also that confined EM waves (*e.g.* for propagation inside of wave guides) are not (in general) purely transverse waves!

The boundary conditions (1)  $\tilde{E}^{\parallel} = 0$  and 2)  $\tilde{B}^{\perp} = 0$  will (in general, for confined waves) require longitudinal components:  $\tilde{E}_{o_z}(x, y)$  and  $\tilde{B}_{o_z}(x, y)$ . Generically, our  $\tilde{\vec{E}}$  and  $\tilde{\vec{B}}$ - fields interior to the wave guide will thus be of the form(s):

$$\begin{aligned}\tilde{\vec{E}}(x, y, z, t) &= \tilde{\vec{E}}_o(x, y) e^{i(k_z z - \omega t)} & \text{with: } & \tilde{\vec{E}}_o(x, y) = \tilde{E}_{o_x}(x, y) \hat{x} + \tilde{E}_{o_y}(x, y) \hat{y} + \tilde{E}_{o_z}(x, y) \hat{z} \\ \text{and: } \tilde{\vec{B}}(x, y, z, t) &= \tilde{\vec{B}}_o(x, y) e^{i(k_z z - \omega t)} & \text{with: } & \tilde{\vec{B}}_o(x, y) = \tilde{B}_{o_x}(x, y) \hat{x} + \tilde{B}_{o_y}(x, y) \hat{y} + \tilde{B}_{o_z}(x, y) \hat{z}\end{aligned}$$

If these expressions are inserted into (3) Faraday's Law and (4) Ampere's Law (above) we obtain:

$$\begin{aligned}(3) \text{ Faraday's Law:} & & (4) \text{ Ampere's Law:} & \\ \text{(i)} & \frac{\partial \tilde{E}_{o_y}}{\partial x} - \frac{\partial \tilde{E}_{o_x}}{\partial y} = i\omega \tilde{B}_{o_z} & \text{(iv)} & \frac{\partial \tilde{B}_{o_y}}{\partial x} - \frac{\partial \tilde{B}_{o_x}}{\partial y} = -\frac{i\omega}{c^2} \tilde{E}_{o_z} \\ \text{(ii)} & \frac{\partial \tilde{E}_{o_z}}{\partial y} - \frac{\partial \tilde{E}_{o_y}}{\partial z} = i\omega \tilde{B}_{o_x} & \text{(v)} & \frac{\partial \tilde{B}_{o_z}}{\partial y} - \frac{\partial \tilde{B}_{o_y}}{\partial z} = -\frac{i\omega}{c^2} \tilde{E}_{o_x} \\ & \frac{\partial \tilde{E}_{o_z}}{\partial y} - ik_z \tilde{E}_{o_y} = i\omega \tilde{B}_{o_x} & & \frac{\partial \tilde{B}_{o_z}}{\partial y} - ik_z \tilde{B}_{o_y} = -\frac{i\omega}{c^2} \tilde{E}_{o_x} \\ \text{(iii)} & \frac{\partial \tilde{E}_{o_x}}{\partial z} - \frac{\partial \tilde{E}_{o_z}}{\partial x} = i\omega \tilde{B}_{o_y} & \text{(vi)} & \frac{\partial \tilde{B}_{o_x}}{\partial z} - \frac{\partial \tilde{B}_{o_z}}{\partial x} = -i\frac{\omega}{c^2} \tilde{E}_{o_y} \\ & ik_z \tilde{E}_{o_x} - \frac{\partial \tilde{E}_{o_z}}{\partial x} = i\omega \tilde{B}_{o_y} & & ik_z \tilde{B}_{o_x} - \frac{\partial \tilde{B}_{o_z}}{\partial x} = -i\frac{\omega}{c^2} \tilde{E}_{o_y}\end{aligned}$$

Note the cyclic permutations in  $x, y, z$  for (i)-(iii) and (iv)-(vi).

We can use the four equations (ii), (iii), (v), and (vi) to solve for  $\tilde{E}_{o_x}$ ,  $\tilde{E}_{o_y}$ ,  $\tilde{B}_{o_x}$  and  $\tilde{B}_{o_y}$  in terms of  $\tilde{E}_{o_z}$  and  $\tilde{B}_{o_z}$ , which, after some algebra yield:

$$\begin{aligned}
 \text{(a)} \quad \tilde{E}_{o_x} &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{E}_{o_z}}{\partial x} + \omega \frac{\partial \tilde{B}_{o_z}}{\partial y} \right) \\
 \text{(b)} \quad \tilde{E}_{o_y} &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{E}_{o_z}}{\partial y} - \omega \frac{\partial \tilde{B}_{o_z}}{\partial x} \right) \\
 \text{(c)} \quad \tilde{B}_{o_x} &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{B}_{o_z}}{\partial x} - \omega \frac{\partial \tilde{E}_{o_z}}{\partial y} \right) \\
 \text{(d)} \quad \tilde{B}_{o_y} &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{B}_{o_z}}{\partial y} + \omega \frac{\partial \tilde{E}_{o_z}}{\partial x} \right)
 \end{aligned}$$

We now insert (a) – (d) above into the other two Maxwell's equations:

(1) Gauss' Law  $\vec{\nabla} \cdot \vec{\tilde{E}} = 0$  and (2) No magnetic charges  $\vec{\nabla} \cdot \vec{\tilde{B}} = 0$ :

$$\frac{\partial \tilde{E}_{o_x}}{\partial x} + \frac{\partial \tilde{E}_{o_y}}{\partial y} + \frac{\partial \tilde{E}_{o_z}}{\partial z} = 0 \quad \text{and:} \quad \frac{\partial \tilde{B}_{o_x}}{\partial x} + \frac{\partial \tilde{B}_{o_y}}{\partial y} + \frac{\partial \tilde{B}_{o_z}}{\partial z} = 0$$

We obtain (after some more algebra): two decoupled wave equations for  $\tilde{E}_{o_z}$  and  $\tilde{B}_{o_z}$ :

$$\begin{aligned}
 \text{(}\alpha\text{)} \quad & \left[ \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \left( \frac{\omega}{c} \right)^2 - k_z^2 \right] \tilde{E}_{o_z} = 0 \\
 \text{(}\beta\text{)} \quad & \left[ \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \left( \frac{\omega}{c} \right)^2 - k_z^2 \right] \tilde{B}_{o_z} = 0
 \end{aligned}$$

For monochromatic plane *EM* traveling waves propagating in the  $+\hat{z}$  direction:

If:  $\tilde{E}_{o_z} = 0$ , these *EM* waves correspond to **TE** (Transverse Electric) waves.

If:  $\tilde{B}_{o_z} = 0$ , these *EM* waves correspond to **TM** (Transverse Magnetic) waves.

If **both**  $\tilde{E}_{o_z} = \tilde{B}_{o_z} = 0$ , these *EM* waves correspond to **TEM** (Transverse Electric & Magnetic) waves.

*n.b.* **TEM** waves cannot propagate in hollow wave guides.  
 {they can propagate *e.g.* in a coaxial waveguide structure with a center conductor}.

If  $\boxed{\tilde{E}_{o_z} = 0}$  {TE waves}, then Gauss' Law ( $\vec{\nabla} \cdot \tilde{\vec{E}} = 0$ ) becomes:

$$\frac{\partial \tilde{E}_{o_x}}{\partial x} + \frac{\partial \tilde{E}_{o_y}}{\partial y} = 0$$

If  $\boxed{\tilde{B}_{o_z} = 0}$  {TM waves}, then Faraday's Law ( $\vec{\nabla} \times \tilde{\vec{E}} = -\frac{\partial \tilde{B}_z}{\partial t}$ ) becomes:

$$\frac{\partial \tilde{E}_{o_y}}{\partial x} - \frac{\partial \tilde{E}_{o_x}}{\partial y} = 0$$

If **both**  $\boxed{\tilde{E}_{o_z} = \tilde{B}_{o_z} = 0}$  {TEM waves}, from ( $\alpha$ ) and ( $\beta$ ) above, we see that  $k = \omega/c$ .  
 $\Rightarrow$  we must go back and fully solve equations (i) – (vi) on page 2 (above).

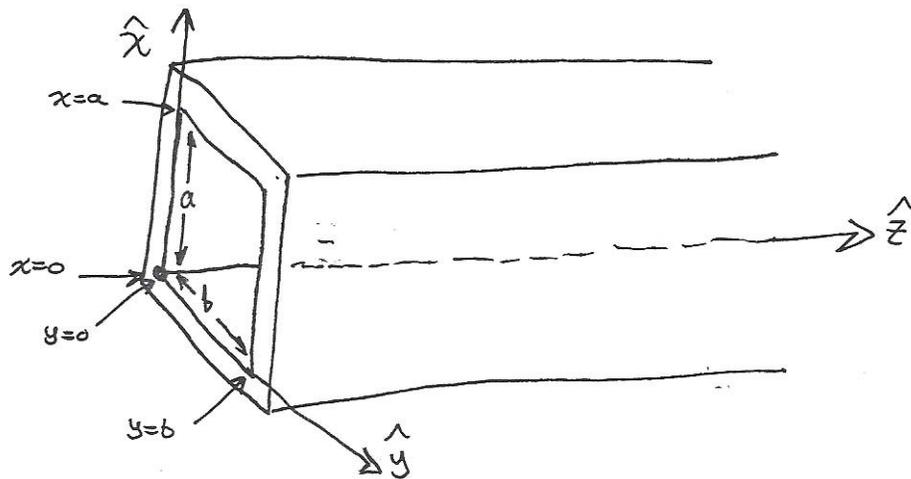
Note that  $\tilde{\vec{E}}_o$  does satisfy  $\vec{\nabla} \cdot \tilde{\vec{E}} = 0$  and  $\vec{\nabla} \times \tilde{\vec{E}} = 0$  (i.e.  $\tilde{\vec{E}}_o$  has zero divergence and zero curl)  
 $\Rightarrow \tilde{\vec{E}}_o = -\vec{\nabla} \tilde{V}_{scalar} \Rightarrow$  then  $\tilde{V}_{scalar}$  satisfies Laplace's equation:  $\vec{\nabla} \cdot (-\vec{\nabla} \tilde{V}) = -\nabla^2 \tilde{V} = 0$

But boundary condition (1):  $\boxed{\tilde{E}^{\parallel} = 0}$  at the inner surface of waveguide  $\Rightarrow$  the inner surface of the waveguide is an equipotential, i.e.  $\tilde{V} = \text{constant}$  on the inner surface of the wave guide.

If the inside of the waveguide is completely hollow, since Laplace's equation does not allow local maxima or minima anywhere except on the surfaces, then {here} the potential  $\tilde{V}$  interior to this wave guide must be a constant everywhere  $\Rightarrow \tilde{\vec{E}}_o = -\vec{\nabla} \tilde{V} = 0$  everywhere inside the waveguide.  $\Rightarrow$  No TEM wave propagation can occur in hollow wave guides {\*unless the wavelength  $\lambda \ll$  cross-sectional dimensions  $a, b$  of the waveguide – then TEM waves are a special / limiting case of TE waves... e.g. EM light waves in an optical fiber = waveguide!!!}.

### **Propagation of TE Waves in a Perfectly Conducting Hollow Rectangular Waveguide** ( $\sigma_c = \infty$ ):

Consider a perfectly conducting, hollow rectangular waveguide of (inner) height  $a$  and width  $b$  as shown in the figure below {n.b. important:  $a \geq b$  by convention!!!}:



For TE waves:  $\tilde{E}_{o_z}(x, y) = 0$ , then:

$$(\alpha) \left[ \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \left(\frac{\omega}{c}\right)^2 - k_z^2 \right] \tilde{E}_{o_z} = 0 \Rightarrow \underline{0=0} \text{ (i.e. no information).}$$

$$(\beta) \left[ \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \left(\frac{\omega}{c}\right)^2 - k_z^2 \right] \tilde{B}_{o_z} = 0 \Rightarrow \tilde{B}_{o_z}(x, y) = 0 \text{ for TE waves.}$$

The boundary condition for  $\tilde{B}_o(x, y)$  is  $\tilde{B}^\perp = 0$  on the inner walls of waveguide.

But:  $\tilde{B}_o(x, y) = \tilde{B}_{o_x}(x, y)\hat{x} + \tilde{B}_{o_y}(x, y)\hat{y} + \tilde{B}_{o_z}(x, y)\hat{z}$ . Then, referring to the above figure:

$$\tilde{B}^\perp = 0 \text{ in the } \hat{x} \text{-direction: } \tilde{B}_{o_x}(x=0, y) = \tilde{B}_{o_x}(x=a, y) = 0$$

$$\tilde{B}^\perp = 0 \text{ in the } \hat{y} \text{-direction: } \tilde{B}_{o_y}(x, y=0) = \tilde{B}_{o_y}(x, y=b) = 0$$

But from equations (c) and (d) above:

$$(c) \tilde{B}_{o_x}(x, y) = \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial x} - \omega \frac{\partial \tilde{E}_{o_z}(x, y)}{\partial y} \right)$$

$$\text{then: } \tilde{B}_{o_x}(x=0, y) = \tilde{B}_{o_x}(x=a, y) = 0 \Rightarrow \frac{\partial \tilde{B}_{o_z}(x=0, y)}{\partial x} = \frac{\partial \tilde{B}_{o_z}(x=a, y)}{\partial x} = 0$$

**n.b.** These terms = 0 because  $E_{o_z}(x, y) = 0$  for TE waves.

$$(d) \tilde{B}_{o_y}(x, y) = \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial y} + \omega \frac{\partial \tilde{E}_{o_z}(x, y)}{\partial x} \right)$$

$$\text{then: } \tilde{B}_{o_y}(x, y=0) = \tilde{B}_{o_y}(x, y=b) = 0 \Rightarrow \frac{\partial \tilde{B}_{o_z}(x, y=0)}{\partial y} = \frac{\partial \tilde{B}_{o_z}(x, y=b)}{\partial y} = 0$$

Now, to solve the wave equation for  $\tilde{B}_{o_z}(x, y)$ :

$$\text{Namely } (\beta) \left[ \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \left(\frac{\omega}{c}\right)^2 - k_z^2 \right] \tilde{B}_{o_z}(x, y) = 0$$

Use the separation of variables technique – try a product solution of the form:  $\tilde{B}_{o_z}(x, y) = \tilde{X}(x)\tilde{Y}(y)$

$$\text{Inserting this into the above equation } (\beta): \tilde{Y}(y) \frac{\partial^2 \tilde{X}(x)}{\partial x^2} + \tilde{X}(x) \frac{\partial^2 \tilde{Y}(y)}{\partial y^2} + \left[ \left(\frac{\omega}{c}\right)^2 - k_z^2 \right] \tilde{X}(x)\tilde{Y}(y) = 0$$

Divide through by  $\tilde{X}(x)\tilde{Y}(y)$ :

$$\underbrace{\frac{1}{\tilde{X}(x)} \frac{\partial^2 \tilde{X}(x)}{\partial x^2}}_{\text{fcn of } x \text{ only}} + \underbrace{\frac{1}{\tilde{Y}(y)} \frac{\partial^2 \tilde{Y}(y)}{\partial y^2}}_{\text{fcn of } y \text{ only}} = - \left[ \left(\frac{\omega}{c}\right)^2 - k_z^2 \right] = \text{constant}$$

The above relation can be true for arbitrary (x,y) points **iff** (if and only if):

$$\begin{aligned}
 (\gamma) \quad & \boxed{\left( \frac{1}{\tilde{X}(x)} \frac{\partial^2 \tilde{X}(x)}{\partial x^2} \right) = -k_x^2 = \text{constant}} \\
 (\delta) \quad & \boxed{\left( \frac{1}{\tilde{Y}(y)} \frac{\partial^2 \tilde{Y}(y)}{\partial y^2} \right) = -k_y^2 = \text{constant}' (\neq -k_x^2)}
 \end{aligned}$$

Then the so-called characteristic equation becomes:

$$\boxed{-k_x^2 - k_y^2 = - \left[ \left( \frac{\omega}{c} \right)^2 - k_z^2 \right] = \text{constant}''} \quad \text{or:} \quad \boxed{k_z^2(\omega) = \left( \frac{\omega}{c} \right)^2 - k_x^2 - k_y^2} \quad \leftarrow \quad \boxed{\text{n.b. } k_z(\omega) \text{ is frequency dependent!}}$$

We can rewrite the characteristic equation as:  $\boxed{\left( \frac{\omega}{c} \right)^2 = k_x^2 + k_y^2 + k_z^2(\omega) = k^2 = |\vec{k}|^2 = \vec{k} \cdot \vec{k}}$

The general solutions of the equations  $\boxed{\frac{\partial^2 \tilde{X}(x)}{\partial x^2} + k_x^2 \tilde{X}(x) = 0}$  and:  $\boxed{\frac{\partial^2 \tilde{Y}(y)}{\partial y^2} + k_y^2 \tilde{Y}(y) = 0}$  are of the form:  $\boxed{\tilde{X}(x) = \tilde{A}_x \cos(k_x x) + \tilde{B}_x \sin(k_x x)}$  and:  $\boxed{\tilde{Y}(y) = \tilde{A}_y \cos(k_y y) + \tilde{B}_y \sin(k_y y)}$

Now the boundary condition  $\boxed{\tilde{B}^\perp = 0}$  requires not only:

$$\text{(c):} \quad \boxed{\tilde{B}_{o_x}(x=0, y) = \tilde{B}_{o_x}(x=a, y) = 0} \quad \text{but also:} \quad \boxed{\frac{\partial \tilde{B}_{o_z}(x=0, y)}{\partial x} = \frac{\partial \tilde{B}_{o_z}(x=a, y)}{\partial x} = 0}$$

$$\text{(d):} \quad \boxed{\tilde{B}_{o_y}(x, y=0) = \tilde{B}_{o_y}(x, y=b) = 0} \quad \text{but also:} \quad \boxed{\frac{\partial \tilde{B}_{o_z}(x, y=0)}{\partial y} = \frac{\partial \tilde{B}_{o_z}(x, y=b)}{\partial y} = 0}$$

Since  $\tilde{B}_{o_x}(x, y) = \tilde{X}(x)\tilde{Y}(y)$  these LATTER boundary conditions require:

$$\boxed{\frac{\partial \tilde{X}(x=0)}{\partial x} = \frac{\partial \tilde{X}(x=a)}{\partial x} = 0} \quad \text{and} \quad \boxed{\frac{\partial \tilde{Y}(y=0)}{\partial y} = \frac{\partial \tilde{Y}(y=b)}{\partial y} = 0}$$

So if:  $\boxed{\tilde{X}(x) = \tilde{A}_x \cos(k_x x) + \tilde{B}_x \sin(k_x x)}$  and  $\boxed{\tilde{Y}(y) = \tilde{A}_y \cos(k_y y) + \tilde{B}_y \sin(k_y y)}$

Then:  $\boxed{\frac{\partial \tilde{X}(x)}{\partial x} = -k_x \tilde{A}_x \sin(k_x x) + k_x \tilde{B}_x \cos(k_x x)}$  and  $\boxed{\frac{\partial \tilde{Y}(y)}{\partial y} = -k_y \tilde{A}_y \sin(k_y y) + k_y \tilde{B}_y \cos(k_y y)}$

Thus:  $\boxed{\frac{\partial \tilde{X}(x=0)}{\partial x} = 0}$  requires:  $\boxed{\tilde{B}_x = 0}$  and  $\boxed{\frac{\partial \tilde{Y}(y=0)}{\partial y} = 0}$  requires:  $\boxed{\tilde{B}_y = 0}$

Likewise:

$$\frac{\partial \tilde{X}(x=a)}{\partial x} = 0 \text{ requires: } k_x a = m\pi, m = 0, 1, 2, 3, \dots \text{ or: } k_x = \left(\frac{m\pi}{a}\right), m = 0, 1, 2, 3, \dots$$

and:

$$\frac{\partial Y(y=b)}{\partial y} = 0 \text{ requires: } k_y b = n\pi, n = 0, 1, 2, 3, \dots \text{ or: } k_y = \left(\frac{n\pi}{b}\right), n = 0, 1, 2, 3, \dots$$

Then  $\tilde{B}_{o_z}(x, y) = \tilde{X}(x)\tilde{Y}(y)$  becomes, after absorbing the coefficients  $\tilde{A}_x$  &  $\tilde{A}_y$  into a single coefficient  $\tilde{B}_o$ :

$$\tilde{B}_{o_z}(x, y) = \tilde{B}_o \cos\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right) \quad \begin{matrix} m = 0, 1, 2, 3, \dots \\ n = 0, 1, 2, 3, \dots \end{matrix}$$

The full  $(x, y, z, t)$  dependence is:  $\tilde{B}_z(x, y, z, t) = \tilde{B}_{o_z}(x, y)e^{i(k_z z - \omega t)}$

The characteristic equation then becomes:

$$k_z^2(\omega) = \left(\frac{\omega}{c}\right)^2 - k_x^2 - k_y^2 = \left(\frac{\omega}{c}\right)^2 - \left(\frac{m\pi}{a}\right)^2 - \left(\frac{n\pi}{b}\right)^2 \quad \begin{matrix} m = 0, 1, 2, 3, \dots \\ n = 0, 1, 2, 3, \dots \end{matrix}$$

Thus, having found/determined  $\tilde{B}_{o_z}(x, y)$  and, since for the TE mode  $\tilde{E}_{o_z}(x, y) \equiv 0$ , we can now determine  $\tilde{E}_{o_x}$ ,  $\tilde{E}_{o_y}$ ,  $\tilde{B}_{o_x}$ , and  $\tilde{B}_{o_y}$  using equations (a) – (d) above:

$$\begin{aligned} \text{(a)} \quad \tilde{E}_{o_x}(x, y) &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{E}_{o_z}(x, y)}{\partial x} + \omega \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial y} \right) = \frac{i\omega}{(\omega/c)^2 - k_z^2} \left( \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial y} \right) \\ \text{(b)} \quad \tilde{E}_{o_y}(x, y) &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{E}_{o_z}(x, y)}{\partial y} - \omega \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial x} \right) = \frac{-i\omega}{(\omega/c)^2 - k_z^2} \left( \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial x} \right) \\ \text{(c)} \quad \tilde{B}_{o_x}(x, y) &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial x} - \omega \frac{\partial \tilde{E}_{o_z}(x, y)}{\partial y} \right) = \frac{ik}{(\omega/c)^2 - k_z^2} \left( \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial x} \right) \\ \text{(d)} \quad \tilde{B}_{o_y}(x, y) &= \frac{i}{(\omega/c)^2 - k_z^2} \left( k_z \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial y} + \omega \frac{\partial \tilde{E}_{o_z}(x, y)}{\partial x} \right) = \frac{ik}{(\omega/c)^2 - k_z^2} \left( \frac{\partial \tilde{B}_{o_z}(x, y)}{\partial y} \right) \end{aligned}$$

But:  $\tilde{B}_{o_z}(x, y) = \tilde{B}_o \cos(k_x x) \cos(k_y y)$  with  $k_x = \left(\frac{m\pi}{a}\right)$ ,  $k_y = \left(\frac{n\pi}{b}\right)$  and  $\begin{matrix} m = 0, 1, 2, 3, \dots \\ n = 0, 1, 2, 3, \dots \end{matrix}$

Explicitly carrying out the spatial differentiation in (a)-(d) above, then for TE wave propagation:

$$\begin{aligned}
 \text{(a)} \quad \tilde{E}_{o_x}(x, y) &= \frac{-i\omega k_y}{(\omega/c)^2 - k_z^2} \tilde{B}_o \cos(k_x x) \sin(k_y y) \quad \text{with} \quad k_x = \left(\frac{m\pi}{a}\right), \quad k_y = \left(\frac{n\pi}{b}\right), \quad \begin{matrix} m = 0, 1, 2, 3, \dots \\ n = 0, 1, 2, 3, \dots \end{matrix} \\
 \text{(b)} \quad \tilde{E}_{o_y}(x, y) &= \frac{+i\omega k_x}{(\omega/c)^2 - k_z^2} \tilde{B}_o \sin(k_x x) \cos(k_y y) \quad \text{and:} \quad k_z^2 = \left(\frac{\omega}{c}\right)^2 - k_x^2 - k_y^2 \quad \text{and:} \\
 \text{(c)} \quad \tilde{E}_{o_z}(x, y) &\equiv 0 \\
 \text{(d)} \quad \tilde{B}_{o_x}(x, y) &= \frac{-ik_z k_x}{(\omega/c)^2 - k_z^2} \tilde{B}_o \sin(k_x x) \cos(k_y y) \\
 \text{(e)} \quad \tilde{B}_{o_y}(x, y) &= \frac{-ik_z k_y}{(\omega/c)^2 - k_z^2} \tilde{B}_o \cos(k_x x) \sin(k_y y) \quad \text{and:} \\
 \text{(f)} \quad \tilde{B}_{o_z}(x, y) &= \tilde{B}_o \cos(k_x x) \cos(k_y y)
 \end{aligned}$$

The full  $(x, y, z, t)$  – dependence is:

$$\begin{aligned}
 \tilde{\vec{E}} = \tilde{E}_x \hat{x} + \tilde{E}_y \hat{y} + \tilde{E}_z \hat{z} \quad & \left\{ \begin{aligned}
 \text{(a)} \quad \tilde{E}_x(x, y, z, t) &= \tilde{E}_{o_x}(x, y) e^{i(k_z z - \omega t)} = \frac{-i\omega k_y}{(\omega/c)^2 - k_z^2} B_o \cos(k_x x) \sin(k_y y) e^{i(k_z z - \omega t)} \\
 \text{(b)} \quad \tilde{E}_y(x, y, z, t) &= \tilde{E}_{o_y}(x, y) e^{i(k_z z - \omega t)} = \frac{+i\omega k_x}{(\omega/c)^2 - k_z^2} B_o \sin(k_x x) \cos(k_y y) e^{i(k_z z - \omega t)} \\
 \text{(c)} \quad \tilde{E}_z(x, y, z, t) &= \tilde{E}_{o_z}(x, y) e^{i(k_z z - \omega t)} = 0
 \end{aligned} \right. \\
 \tilde{\vec{B}} = \tilde{B}_x \hat{x} + \tilde{B}_y \hat{y} + \tilde{B}_z \hat{z} \quad & \left\{ \begin{aligned}
 \text{(d)} \quad \tilde{B}_x(x, y, z, t) &= \tilde{B}_{o_x}(x, y) e^{i(k_z z - \omega t)} = \frac{-ik_z k_x}{(\omega/c)^2 - k_z^2} B_o \sin(k_x x) \cos(k_y y) e^{i(k_z z - \omega t)} \\
 \text{(e)} \quad \tilde{B}_y(x, y, z, t) &= \tilde{B}_{o_y}(x, y) e^{i(k_z z - \omega t)} = \frac{-ik_z k_y}{(\omega/c)^2 - k_z^2} B_o \cos(k_x x) \sin(k_y y) e^{i(k_z z - \omega t)} \\
 \text{(f)} \quad \tilde{B}_z(x, y, z, t) &= \tilde{B}_{o_z}(x, y) e^{i(k_z z - \omega t)} = B_o \cos(k_x x) \cos(k_y y) e^{i(k_z z - \omega t)}
 \end{aligned} \right.
 \end{aligned}$$

Note that for the TE mode(s) of propagation of *EM* waves in a rectangular waveguide, the  $\tilde{\vec{E}}$  and  $\tilde{\vec{B}}$  -fields are in phase with each other – the  $x$ ,  $y$  and  $z$ -components of  $\tilde{\vec{E}}$  and  $\tilde{\vec{B}}$  all have the common phase factor  $e^{i(k_z z - \omega t)}$ .

The wave number  $k_z(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - k_x^2 - k_y^2} = \frac{2\pi}{\lambda_z}$  with  $k_x = \left(\frac{m\pi}{a}\right)$ ,  $k_y = \left(\frac{n\pi}{b}\right)$  and  $m = 0, 1, 2, 3, \dots$   
 $n = 0, 1, 2, 3, \dots$

Thus:  $k_z(\omega) = \frac{2\pi}{\lambda_z(\omega)} = \sqrt{\left(\frac{\omega}{c}\right)^2 - [k_x^2 + k_y^2]} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]}$

We can define a so-called {angular} cutoff frequency for the  $(m,n)^{th}$  TE mode as:

$$\omega_{m,n} \equiv c \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

Thus, we can rewrite the above relation as:

$$k_z^{m,n}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left(\frac{\omega_{m,n}}{c}\right)^2} = \frac{1}{c} \sqrt{\omega^2 - \omega_{m,n}^2}$$

Note that for {angular} frequencies below the cutoff frequency:  $\omega < \omega_{m,n} \equiv c \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$

Then:  $(\omega^2 - \omega_{m,n}^2) < 0$  and:  $k_z^{m,n}(\omega) = \frac{1}{c} \sqrt{\omega^2 - \omega_{m,n}^2}$  becomes imaginary, hence:  $e^{i(k_z z)} \rightarrow e^{-k_z z}$   
 which means that when  $\omega < \omega_{m,n}$ , the EM wave for the  $(m,n)^{th}$  mode is exponentially damped.

Note also that  $m = n = 0$  corresponds to  $k_x = k_y = 0$  with  $k_z^{0,0}(\omega) = \omega/c$ . But then, from the above  $\tilde{E}$ - and  $\tilde{B}$ -field relations on the previous page, we see that for this kind of TE wave, that:

$$\tilde{E}_x = \tilde{E}_y = \tilde{E}_z = 0 \quad \text{and} \quad \tilde{B}_x = \tilde{B}_y = 0 \quad \text{with} \quad \tilde{B}_z = B_0 e^{i(k_z z - \omega t)} = 0.$$

$\Rightarrow$  This is not a proper kind of propagating EM wave, because then  $\tilde{E} = \tilde{B} = 0$  everywhere!!!

Thus, the lowest non-trivial propagating TE-type EM wave is the TE<sub>10</sub> mode, where the notation TE<sub>mn</sub> designates the  $(m,n)^{th}$  mode of propagation. Note again, that by convention, the index associated with the largest transverse dimension (here  $a$ ) with corresponding integer index  $m$  is given first.

Thus, for the lowest TE mode, TE<sub>1,0</sub>:  $k_z^{1,0}(\omega) = \frac{1}{c} \sqrt{\omega^2 - \omega_{1,0}^2} = \frac{1}{c} \sqrt{\omega^2 - \left(\frac{\pi c}{a}\right)^2}$

and we see that  $k_{1,0} \geq 0$  {i.e. is a purely real quantity} when:  $\omega^2 - \omega_{1,0}^2 = \omega^2 - (\pi c/a)^2 > 0$   
 i.e. when:  $\omega > \omega_{1,0} \equiv (\pi c/a)$  (radians/sec), or:  $f > f_{1,0} = (\omega_{1,0}/2\pi) = c/2a$  (Hz).

### A Numerical Example:

Suppose the rectangular wave guide's transverse internal dimensions are  $a = 2 \text{ cm}$  and  $b = 1 \text{ cm}$

Then:  $\omega_{1,0} = \pi c/a = 3\pi \times 10^8 \text{ m/s} / 0.02 \text{ m} = 1.5\pi \times 10^{10} \text{ radians/sec} = 4.71 \times 10^{10} \text{ radians/sec}$

This corresponds to a cutoff frequency of:  $f_{1,0} = \omega_{1,0}/2\pi = \frac{3}{4} \times 10^{10} / \text{s} = 7.5 \text{ GHz}$

which is in the microwave portion of the *EM* spectrum, and corresponds to a wavelength of:

$$\lambda_z^{1,0} = c/f_{1,0} = 3 \times 10^8 \text{ m/s} / \frac{3}{4} \times 10^{10} / \text{s} = 4 \times 10^{-2} \text{ m} = 4.0 \text{ cm} \quad \text{i.e.} \quad \lambda_z^{1,0} = 2a \quad !!!$$

Thus, we see that if  $\lambda_z > \lambda_z^{1,0} = 2a$ , we **cannot** propagate  $\text{TE}_{1,0}$  waves because:  $f < f_{1,0} = \frac{c}{\lambda_z^{1,0}} = \frac{c}{2a}$ .

We also see that if  $\lambda_z < \lambda_z^{1,0} = 2a$ , then we **can** propagate  $\text{TE}_{1,0}$  wave because:  $f > f_{1,0} = \frac{c}{\lambda_z^{1,0}} = \frac{c}{2a}$ .

Precisely at the angular cutoff frequency for the  $\text{TE}_{1,0}$  mode, i.e.  $\omega = \omega_{1,0} = \pi c/a$ , we see that the wavenumber  $k_z^{1,0}(\omega) = \frac{1}{c} \sqrt{\omega^2 - \omega_{1,0}^2} = 0 = 2\pi/\lambda_z^{1,0}(\omega)$  and thus  $\lambda_z^{1,0}(\omega) = \infty$  for  $\omega = \omega_{1,0}$ , where  $\lambda_z^{1,0}(\omega) =$  the wavelength of the *EM* wave in the waveguide for the  $\text{TE}_{1,0}$  mode.

Now suppose that  $\omega > \omega_{1,0}$  then:  $k_z^{m,n}(\omega) = \frac{2\pi}{\lambda_z^{m,n}(\omega)} = \frac{1}{c} \sqrt{\omega^2 - \omega_{m,n}^2} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]}$

The higher the angular frequency  $\omega$  is, it then becomes possible to propagate  $\text{TE}_{m,n}$  waves in more than just one mode. There exists an angular cutoff frequency for each  $\text{TE}_{m,n}$  mode:

Angular cutoff frequency for each  $\text{TE}_{mn}$  mode:  $\omega_{m,n} \equiv c \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$

**Another Numerical Example:**

A rectangular wave guide's transverse internal dimensions are (again)  $a = 2 \text{ cm}$  and  $b = 1 \text{ cm}$ . Suppose that:  $f = 20 \text{ GHz} = 2 \times 10^{10} \text{ Hz}$ , thus:  $\omega = 2\pi f = 4\pi \times 10^{10} = 12.56 \times 10^{10} \text{ radians/sec}$  with corresponding vacuum wavelength  $\lambda_o = c/f = 1.5 \text{ cm}$ .

Which  $TE_{m,n}$  modes are **accessible**?  $\omega_{m,n} \equiv c\sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$

$$\omega_{1,0} = c\sqrt{\left(\frac{1\pi}{a}\right)^2} = \left(\frac{\pi c}{a}\right) = 4.71 \times 10^{10} \text{ radians/sec}$$

$$\omega_{0,1} = c\sqrt{\left(\frac{1\pi}{b}\right)^2} = \left(\frac{\pi c}{b}\right) = 9.42 \times 10^{10} \text{ radians/sec}$$

$$\omega_{1,1} = c\sqrt{\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2} = \pi c\sqrt{\left(\frac{1}{a}\right)^2 + \left(\frac{1}{b}\right)^2} = 10.53 \times 10^{10} \text{ radians/sec}$$

$$\omega_{2,0} = c\sqrt{\left(\frac{2\pi}{a}\right)^2} = \frac{2\pi c}{a} = 9.42 \times 10^{10} \text{ radians/sec}$$

$$\omega_{3,0} = c\sqrt{\left(\frac{3\pi}{a}\right)^2} = \frac{3\pi c}{a} = 14.14 \times 10^{10} \text{ radians/sec} \leftarrow \text{TOO HIGH !!!}$$

$$\omega_{2,1} = c\sqrt{\left(\frac{2\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2} = \pi c\sqrt{\left(\frac{2}{a}\right)^2 + \left(\frac{1}{b}\right)^2} = \pi c\sqrt{2\left(\frac{1}{b}\right)^2} = 13.33 \times 10^{10} \text{ radians/sec}$$

Thus, for  $f = 20 \text{ GHz} \Rightarrow \omega = 12.56 \times 10^{10} \text{ radians/sec}$  we can access/can propagate  $TE_{m,n}$  waves in the following 4 modes:

$TE_{1,0}:$	$\omega_{1,0} = 4.71 \times 10^{10} \text{ radians/sec}$	
$TE_{0,1}:$	$\omega_{0,1} = 9.42 \times 10^{10} \text{ radians/sec}$	<div style="border: 1px solid black; padding: 5px; display: inline-block;"> <b>n.b.</b> Degenerate,  <u>because</u> <math>a = 2b</math> !!!                 </div>
$TE_{2,0}:$	$\omega_{2,0} = 9.42 \times 10^{10} \text{ radians/sec}$	
$TE_{1,1}:$	$\omega_{1,1} = 10.53 \times 10^{10} \text{ radians/sec}$	

**TE<sub>m,n</sub> Wavenumbers and Wavelengths Inside the Waveguide:**  $a = 2 \text{ cm}$  and  $b = 1 \text{ cm}$ 

$$\left. \begin{array}{l} f = 20 \text{ GHz} \\ = 2 \times 10^{10} \text{ Hz} \\ \omega = 12.56 \times 10^{10} \\ \text{radians/sec} \end{array} \right\} \begin{cases} TE_{1,0} : k_z^{1,0}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left(\frac{\pi}{a}\right)^2} & = 388.31 \text{ m}^{-1}, \lambda_z^{1,0}(\omega) = \frac{2\pi}{k_{1,0}(\omega)} = 1.620 \text{ cm} \\ TE_{0,1} : k_z^{0,1}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left(\frac{\pi}{b}\right)^2} & = 277.06 \text{ m}^{-1}, \lambda_z^{0,1}(\omega) = \frac{\pi}{k_{0,1}(\omega)} = 2.268 \text{ cm} \\ TE_{2,0} : k_z^{2,0}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left(\frac{2\pi}{a}\right)^2} & = 277.06 \text{ m}^{-1}, \lambda_z^{2,0}(\omega) = \frac{\pi}{k_{2,0}(\omega)} = 2.268 \text{ cm} \\ TE_{1,1} : k_z^{1,1}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2\right]} & = 228.23 \text{ m}^{-1}, \lambda_z^{1,1}(\omega) = \frac{\pi}{k_{1,1}(\omega)} = 2.750 \text{ cm} \end{cases}$$

Degenerate !!!

Compare these to vacuum wavenumber  $k_o = \frac{2\pi}{\lambda_o} = 418.88 \text{ m}^{-1}$  and vacuum wavelength  $\lambda_o = 1.5 \text{ cm}$ .

Note that the wavenumbers and wavelengths inside the wave guide will change when the

frequency  $f$  (or  $\omega = 2\pi f$ ) changes, because  $k_z^{m,n}(\omega) = \frac{2\pi}{\lambda_z^{m,n}(\omega)} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]}$

Physically, the phase velocity (also known as the wave velocity)  $v_{\phi_z}^{m,n}(\omega)$  is the speed of propagation of planes of constant phase  $\phi_{m,n}(\omega) \equiv [k_z^{m,n}(\omega)z - \omega t] = \text{constant}$  and is associated with the  $e^{i(k_z z - \omega t)}$  phase-factor of the EM wave for each individual TE<sub>m,n</sub> mode.

Since  $\phi_{m,n}(\omega) \equiv [k_z^{m,n}(\omega)z - \omega t] = \text{constant}$ , then:  $\partial \phi_{m,n}(\omega) / \partial t = 0$  which means that:

$$\frac{\partial \phi_{m,n}(\omega)}{\partial t} = \frac{\partial}{\partial t} [k_z^{m,n}(\omega)z(t) - \omega t] = k_z^{m,n}(\omega) \frac{\partial z(t)}{\partial t} - \omega = 0, \text{ or that: } k_z^{m,n}(\omega) \frac{\partial z(t)}{\partial t} = \omega, \text{ or:}$$

$$\frac{\partial z(t)}{\partial t} = \frac{\omega}{k_z^{m,n}(\omega)}. \text{ The phase velocity } v_{\phi_z}^{m,n}(\omega) \equiv \frac{\partial z(t)}{\partial t} = \frac{\omega}{k_z^{m,n}(\omega)}$$

Thus, the phase velocity of a TE<sub>m,n</sub> wave for the (m,n)<sup>th</sup> mode is:

$$v_{\phi_z}^{m,n}(\omega) \equiv \frac{\omega}{k_z^{m,n}(\omega)} = \frac{\omega}{\sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]}}$$

Since:

$$\omega_{m,n} \equiv c \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} \text{ then we see that the } \underline{\text{phase}} \text{ velocity of a TE}_{m,n} \text{ wave is:}$$

$$v_{\phi_z}^{m,n}(\omega) \equiv \frac{\omega}{k_z^{m,n}(\omega)} = \frac{c}{\sqrt{1 - (\omega_{m,n}/\omega)^2}} > c \text{ for the } (m,n)^{\text{th}} \text{ allowed TE}_{m,n} \text{ mode!!!}$$

For the  $(m,n)^{th}$   $TE_{m,n}$  mode, *EM energy* in the waveguide propagates at the group velocity:

$$v_{g_z}^{m,n}(\omega) \equiv 1 / \left( \frac{dk_z^{m,n}(\omega)}{d\omega} \right) = \left( \frac{dk_z^{m,n}(\omega)}{d\omega} \right)^{-1} \quad \Leftarrow \quad \begin{array}{l} \text{See P436 HW \#7,} \\ \text{Griffiths problem} \\ \text{9.29, page 411} \end{array}$$

Let's calculate  $v_{g_z}^{m,n}(\omega)$ :

$$\frac{dk_z^{m,n}(\omega)}{d\omega} = \frac{d}{d\omega} \left\{ \frac{1}{c} \sqrt{\omega^2 - \omega_{m,n}^2} \right\} = \frac{2\omega}{2c} \cdot \frac{1}{\sqrt{\omega^2 - \omega_{m,n}^2}} = \frac{\omega/c}{\sqrt{\omega^2 - \omega_{m,n}^2}}$$

Thus: 
$$v_{g_z}^{m,n}(\omega) = \frac{1}{\frac{dk_z^{m,n}(\omega)}{d\omega}} = \frac{\omega/c}{\sqrt{\omega^2 - \omega_{m,n}^2}} = c \sqrt{1 - \left( \frac{\omega_{m,n}}{\omega} \right)^2} \quad \text{where: } \omega_{m,n} \equiv c \sqrt{\left( \frac{m\pi}{a} \right)^2 + \left( \frac{n\pi}{b} \right)^2}$$

It can be seen from the above relation that  $v_{g_z}^{m,n}(\omega) < c$  {always!}, as required by causality...

Note further that: 
$$v_{\phi_z}^{m,n}(\omega) \cdot v_{g_z}^{m,n}(\omega) = \frac{c}{\sqrt{1 - (\omega_{m,n}/\omega)^2}} \cdot c \sqrt{1 - (\omega_{m,n}/\omega)^2} = c^2$$

The instantaneous surface charge and current densities induced on the inner surfaces of the {perfectly conducting} waveguide due to the *EM* fields within the waveguide can be obtained from:

$$\sigma_{surf}^{ind}(x, y, z, t) = \epsilon_0 \vec{E}_{surf}(x, y, z, t) \cdot \hat{n}_{surf}(x, y, z)$$

and:

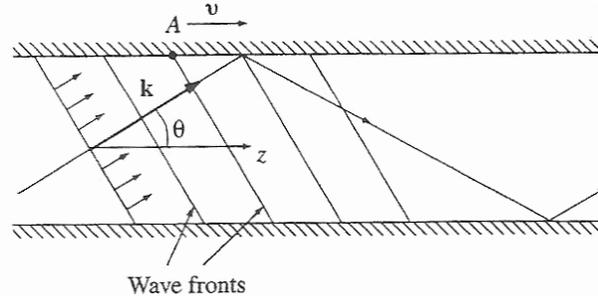
$$\vec{K}_{surf}^{ind}(x, y, z, t) = \frac{1}{\mu_0} \hat{n}_{surf}(x, y, z) \times \vec{B}_{surf}(x, y, z, t)$$

where  $\hat{n}_{surf}(x, y, z)$  is the local {inward-pointing} unit normal at  $(x, y, z)$  associated with a given inner surface of the waveguide, and  $\vec{E}_{surf}(x, y, z, t)$ ,  $\vec{B}_{surf}(x, y, z, t)$  are the instantaneous electric, magnetic fields evaluated at  $(x, y, z, t)$  on that surface.

Note that  $\vec{E}_{surf}(x, y, z, t) \cdot \hat{n}_{surf}(x, y, z)$  is the instantaneous local normal (*i.e.*  $\perp$ ) component of the electric field at  $(x, y, z, t)$  on that surface, whereas  $\hat{n}_{surf}(x, y, z) \times \vec{B}_{surf}(x, y, z, t)$  is the instantaneous local tangential (*i.e.*  $\parallel$ ) component of the magnetic field at  $(x, y, z, t)$  on that surface.

## The Physical Picture of *EM* Waves Propagating Inside a Wave Guide.

Consider an ordinary monochromatic *EM* plane wave initially propagating at speed  $c = \omega/|\vec{k}|$  in the  $\hat{k}$ -direction, making an angle  $\theta$  with respect to the  $\hat{z}$ -axis, as shown in the figure below:



Because the inner walls of the wave guide are perfectly conducting, they are lossless, *i.e.* perfectly reflecting. The *EM* waves are thus multiply-reflected (*n.b.* with  $\pi$  phase shift at each reflection) as they “bounce” down the wave guide – interfering with each other in such a way as to form standing wave patterns of wavelength  $\lambda_x = 2a/m$  in the  $\hat{x}$ -direction and  $\lambda_y = 2b/n$  in the  $\hat{y}$ -direction!!!

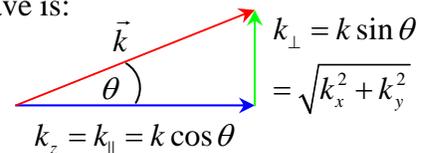
The  $x, y$  wavelengths respectively correspond to the  $x, y$ -wavenumbers  $k_x = 2\pi/\lambda_x = m\pi/a$  in the  $\hat{x}$ -direction, and  $k_y = 2\pi/\lambda_y = n\pi/b$  in the  $\hat{y}$ -direction. In the  $\hat{z}$ -direction, the ensemble (*i.e.* group) of reflected waves results in a traveling wave, with  $z$ -wavenumber:

$$k_z^{m,n}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - (k_x^2 + k_y^2)} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]} = \frac{1}{c} \sqrt{\omega^2 - \omega_{m,n}^2}$$

$$\text{where: } \omega_{m,n} \equiv c \sqrt{(m\pi/a)^2 + (n\pi/b)^2}$$

The propagation wavevector associated with the initial plane wave is:

$$\vec{k} = k_x \hat{x} + k_y \hat{y} + k_z \hat{z} = \left(\frac{m\pi}{a}\right) \hat{x} + \left(\frac{n\pi}{b}\right) \hat{y} + k_z^{m,n}(\omega) \hat{z}$$



Thus, because  $m, n = 0, 1, 2, 3, \dots$  (*n.b.* both  $m = n = 0$  simultaneously is not allowed), only certain angles  $\theta_{m,n}$  will lead to one of the allowed standing wave patterns in  $x$  and  $y$ :

$$\cos \theta_{m,n} = \frac{k_z^{m,n}(\omega)}{|\vec{k}|} = \frac{\sqrt{\omega^2 - \omega_{m,n}^2}/c}{\omega/c} = \sqrt{1 - (\omega_{m,n}/\omega)^2} \quad \text{where: } \omega_{m,n} \equiv c \sqrt{(m\pi/a)^2 + (n\pi/b)^2}$$

This “original” plane *EM* wave, traveling at angle  $\theta_{m,n}$  with respect to the  $\hat{z}$ -axis travels at speed  $c = \omega/|\vec{k}|$  (*i.e.* we assume that the medium (*e.g.* air, or vacuum) inside the wave guide has  $\epsilon = \epsilon_0$  and  $\mu = \mu_0$ ).

But because this plane *EM* wave makes an angle  $\theta_{m,n}$  with respect to the  $\hat{z}$ -axis, the component of the initial wave's speed projected along the  $\hat{z}$ -axis is less than  $c$  :

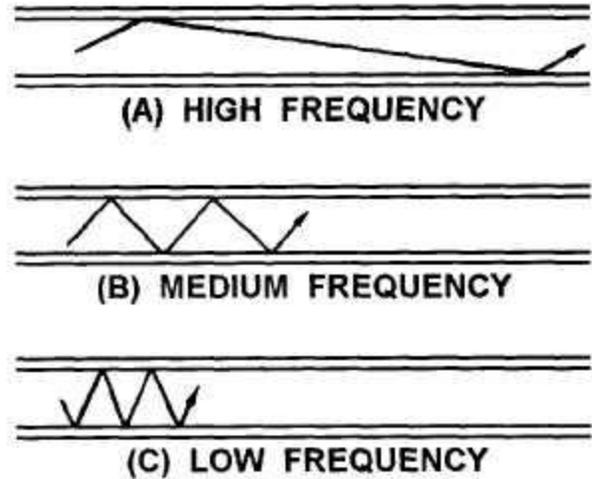
$$v_z(\omega) = c \cos \theta_{m,n}(\omega) = c \sqrt{1 - (\omega_{m,n}/\omega)^2} = v_{g_z}^{m,n}(\omega) = \text{group velocity!}$$

The phase velocity (*aka wave* velocity) is the speed at which wavefronts (planes of constant phase) (*e.g.* point *A* in the figure on the previous page) propagate down the wave guide – these can move much faster than  $c$ , because:

$$v_{\phi_z}^{m,n}(\omega) = \frac{c}{\cos \theta_{m,n}(\omega)} = \frac{c}{\sqrt{1 - (\omega_{m,n}/\omega)^2}}$$

Note that if  $\theta_{m,n} = 90^\circ$  (*i.e.*  $\cos \theta_{m,n} = 0$ ), {*i.e.* when  $\omega = \omega_{m,n}$ }, for which  $v_g^{m,n}(\omega) = 0$  and  $v_{\phi_z}^{m,n}(\omega) = \infty$  !!!

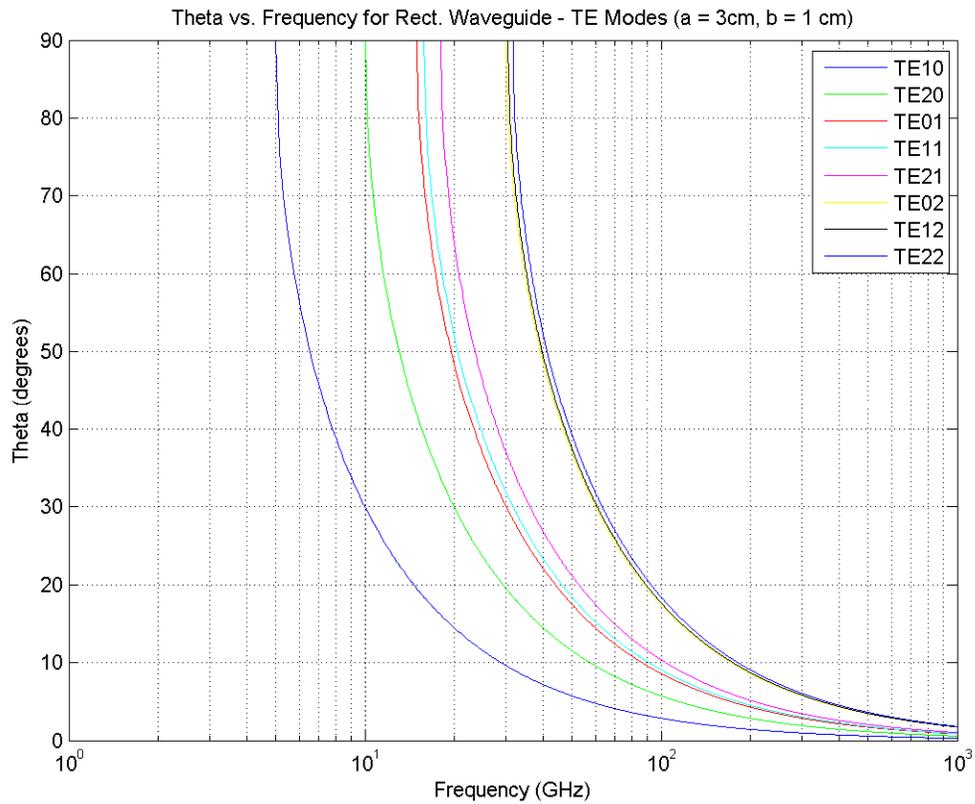
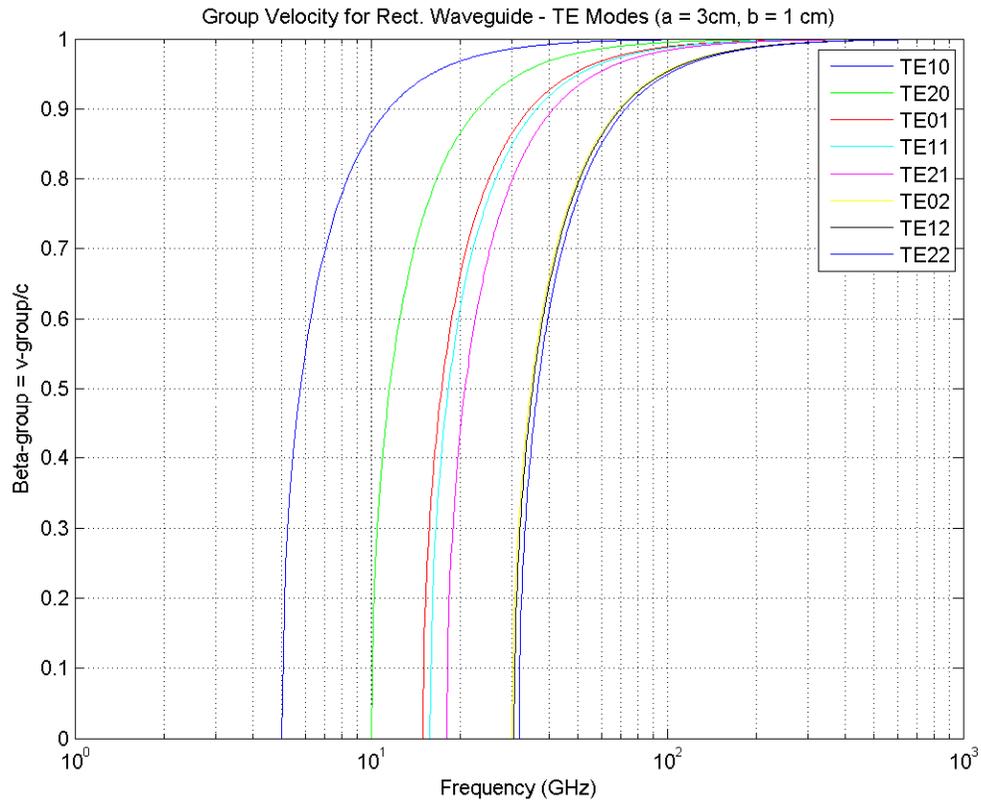
Physically, this corresponds to standing waves in  $(x,y)$ , *i.e.* NO propagation along  $\hat{z}$ -direction  $\Rightarrow$  *i.e.* a 2-D resonant cavity!!!



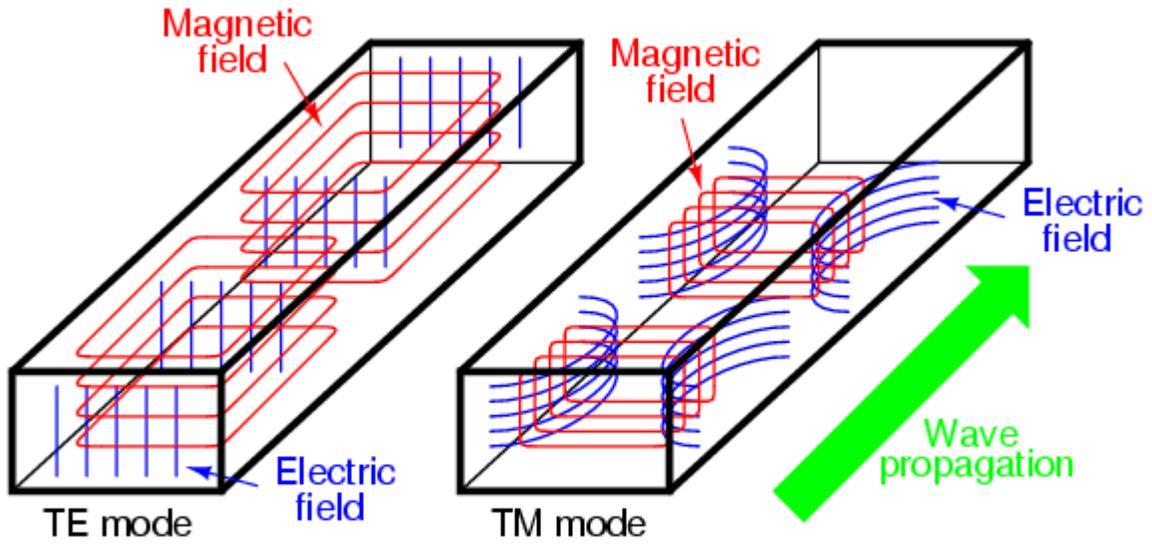
Thus, the allowed solution(s) that we obtained on *p.* 8 above for the  $x$ ,  $y$  and  $z$  components of the electric and magnetic fields for *TE* mode propagation of electromagnetic waves down a waveguide actually/physically represent the steady-state ensemble (*i.e.* group) wave solution associated with the collective effect(s) of the multiply-reflected waves interfering with each other as they propagate down the waveguide!

This group of waves for the  $(m,n^{\text{th}})$  *TE* (and/or *TE*) mode propagates down the waveguide at the group velocity  $v_{g_z}^{m,n}(\omega) = c \sqrt{1 - (\omega_{m,n}/\omega)^2} = c \cos \theta_{m,n}(\omega)$  (hence the origin of its name!).

In the two figures below, we show plots of  $\beta_{g_z}^{m,n}(f) \equiv v_{g_z}^{m,n}(\omega)/c = \sqrt{1 - (f_{m,n}/f)^2}$  vs.  $f$  and  $\theta_{m,n} = \cos^{-1} \left( \beta_{g_z}^{m,n}(f) \equiv v_{g_z}^{m,n}(\omega)/c = \sqrt{1 - (f_{m,n}/f)^2} \right)$  vs.  $f$ :



### 3-D Picture of $\vec{E}$ and $\vec{B}$ -fields in Rectangular Wave Guide for $TE_{1,0}$ Mode:



*Magnetic flux lines appear as continuous loops*  
*Electric flux lines appear with beginning and end points*

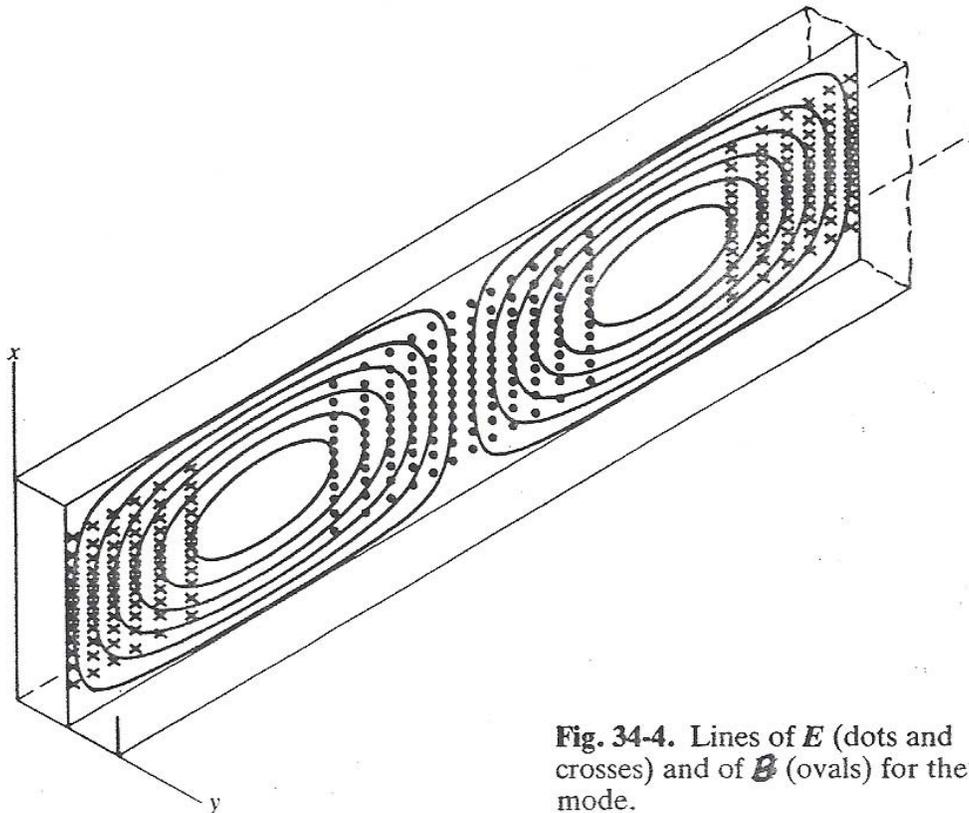
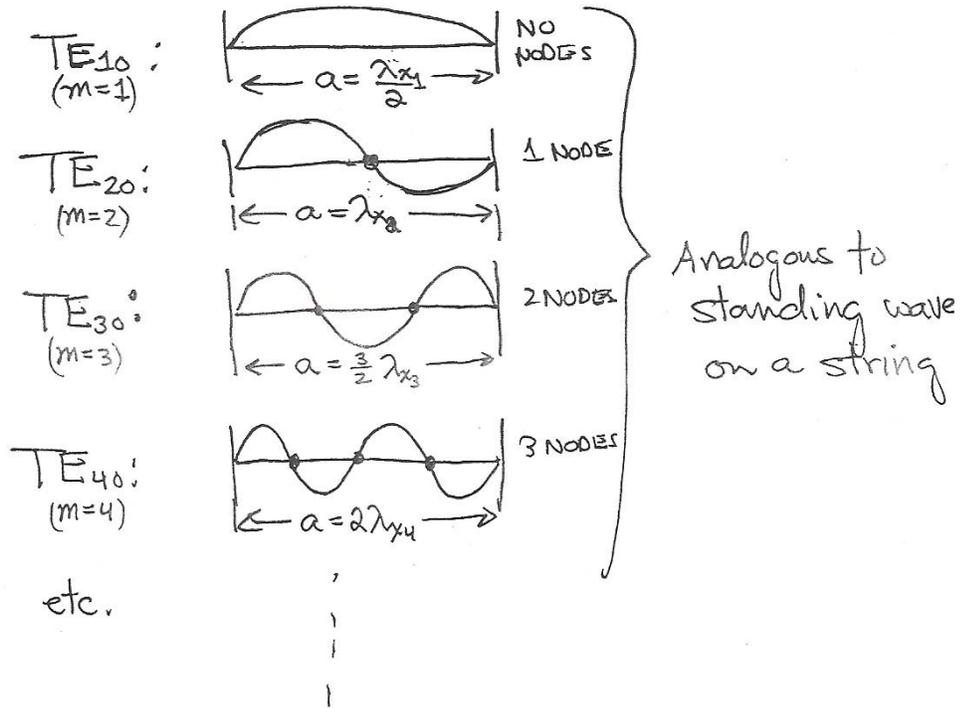


Fig. 34-4. Lines of  $E$  (dots and crosses) and of  $B$  (ovals) for the  $TE_{1,0}$  mode.

For  $TE_{0,1}$  mode, rotate above pix by  $90^\circ$

For  $TE_{m,0}$  modes -  $\exists$  nodes at the mid-plane:



### Time-Averaged Power Transmitted Down a Rectangular Wave Guide in $TE_{m,n}$ Modes:

In order to calculate the time-averaged power transmitted down a rectangular wave guide {of cross-sectional area  $A_{\perp} = ab (= h \times w)$  } we integrate the time-averaged Poynting vector,  $\langle \vec{S}(\vec{r}, t) \rangle_t$  over the cross-sectional area of the waveguide:

$$\langle P_{m,n}^{trans}(z, t) \rangle = \int_{A_{\perp}} \langle \vec{S}_{m,n}(x, y, z, t) \rangle \cdot d\vec{a}_{\perp} = \int_{y=0}^{y=b} \int_{x=0}^{x=a} \langle \vec{S}_{m,n}(x, y, z, t) \rangle \cdot \hat{n} dx dy$$

$\hat{n} = +\hat{z}$  direction (here)

$$d\vec{a}_{\perp} = \hat{n} dx dy = \hat{z} dx dy$$

From Griffiths Problem 9.11 (p.382):

$$\langle \vec{S}_{m,n}(x, y, z, t) \rangle = \frac{1}{2\mu_0} \Re \left( \vec{E}_{m,n}(x, y, z, t) \times \vec{B}_{m,n}^*(x, y, z, t) \right)$$

{Because  $\langle f g \rangle = \frac{1}{2} \Re(\tilde{f} \tilde{g}^*)$ , where \* denotes complex conjugation }

For the  $TE_{m,n}$  modes in a rectangular wave guide:

$$\tilde{\vec{E}}_{m,n}(x, y, z, t) = \tilde{\vec{E}}_{o_{m,n}}(x, y) e^{i(k_z z - \omega t)} \quad \text{and:} \quad \tilde{\vec{B}}_{m,n}^*(x, y, z, t) = \tilde{\vec{B}}_{o_{m,n}}^*(x, y) e^{-i(k_z z - \omega t)} \quad m, n = 0, 1, 2, 3, \dots$$

$$\text{with: } k_z = k_z^{m,n} \equiv \sqrt{\left(\frac{\omega}{c}\right)^2 - [k_{x_m}^2 + k_{y_n}^2]} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]}, \quad k_{x_m} \equiv \left(\frac{m\pi}{a}\right), \quad k_{y_n} \equiv \left(\frac{n\pi}{b}\right)$$

$$\text{and with: } \tilde{\vec{E}}_{o_{m,n}}(x, y) = \tilde{E}_{ox_{mn}} \hat{x} + \tilde{E}_{oy_{mn}} \hat{y} + \tilde{E}_{oz_{mn}} \hat{z} \quad \text{and} \quad \tilde{\vec{B}}_{o_{m,n}}^*(x, y) = \tilde{B}_{ox_{mn}}^* \hat{x} + \tilde{B}_{oy_{mn}}^* \hat{y} + \tilde{B}_{oz_{mn}}^* \hat{z}$$

$$\begin{aligned} \tilde{E}_{ox_{mn}}(x, y) &= \frac{i\omega(-k_{y_n})}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} B_o \cos(k_{x_m} x) \sin(k_{y_n} y) = \frac{i\omega}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} \left(\frac{-n\pi}{b}\right) B_o \cos\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right) \\ \tilde{E}_{oy_{mn}}(x, y) &= \frac{-i\omega(-k_{x_m})}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} B_o \sin(k_{x_m} x) \cos(k_{y_n} y) = \frac{-i\omega}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} \left(\frac{-m\pi}{a}\right) B_o \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right) \\ \tilde{E}_{oz_{mn}}(x, y) &= 0 \\ \tilde{B}_{ox_{mn}}^*(x, y) &= \frac{-ik_{mn}(-k_{x_m})}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} B_o \sin(k_{x_m} x) \cos(k_{y_n} y) = \frac{ik_{mn}}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} \left(\frac{-m\pi}{a}\right) B_o \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right) \\ \tilde{B}_{oy_{mn}}^*(x, y) &= \frac{-ik_{mn}(-k_{y_n})}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} B_o \cos(k_{x_m} x) \sin(k_{y_n} y) = \frac{ik_{mn}}{\left[(\omega/c)^2 - k_{z_{mn}}^2\right]} \left(\frac{-n\pi}{b}\right) B_o \cos\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right) \\ \tilde{B}_{oz_{mn}}^*(x, y) &= B_o \cos(k_{x_m} x) \cos(k_{y_n} y) = B_o \cos\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right) \end{aligned}$$

$$\text{Then: } \langle \vec{S}(x, y, z, t) \rangle = \frac{1}{2\mu_o} \Re \left( \tilde{\vec{E}}(x, y, z, t) \times \tilde{\vec{B}}^*(x, y, z, t) \right) \quad \leftarrow \quad \text{Note: All time dependence vanishes } \{ e^{i(k_z z - \omega t)} \text{ factor} \}$$

Very Useful Table:

$\hat{x} \times \hat{y} = \hat{z}$	$\hat{y} \times \hat{x} = -\hat{z}$	$\hat{x} \times \hat{x} = 0$
$\hat{y} \times \hat{z} = \hat{x}$	$\hat{z} \times \hat{y} = -\hat{x}$	$\hat{y} \times \hat{y} = 0$
$\hat{z} \times \hat{x} = \hat{y}$	$\hat{x} \times \hat{z} = -\hat{y}$	$\hat{z} \times \hat{z} = 0$

$$\begin{aligned} \text{Then: } & \left( \tilde{\vec{E}} = \tilde{E}_x \hat{x} + \tilde{E}_y \hat{y} + \tilde{E}_z \hat{z} \right) \times \left( \tilde{\vec{B}}^* = B_x^* \hat{x} + B_y^* \hat{y} + B_z^* \hat{z} \right) \\ &= \tilde{E}_x \tilde{B}_y^* (\hat{x} \times \hat{y}) + \tilde{E}_x \tilde{B}_z^* (\hat{x} \times \hat{z}) = \tilde{E}_x \tilde{B}_y^* \hat{z} - \tilde{E}_x \tilde{B}_z^* \hat{y} \\ &+ \tilde{E}_y \tilde{B}_x^* (\hat{y} \times \hat{x}) + \tilde{E}_y \tilde{B}_z^* (\hat{y} \times \hat{z}) = -\tilde{E}_y \tilde{B}_x^* \hat{z} + \tilde{E}_y \tilde{B}_z^* \hat{x} \\ &+ \tilde{E}_z \tilde{B}_x^* (\hat{z} \times \hat{x}) + \tilde{E}_z \tilde{B}_y^* (\hat{z} \times \hat{y}) = \tilde{E}_z \tilde{B}_x^* \hat{y} - \tilde{E}_z \tilde{B}_y^* \hat{x} \\ &= (\tilde{E}_y \tilde{B}_z^* - \tilde{E}_z \tilde{B}_y^*) \hat{x} + (\tilde{E}_z \tilde{B}_x^* - \tilde{E}_x \tilde{B}_z^*) \hat{y} + (\tilde{E}_x \tilde{B}_y^* - \tilde{E}_y \tilde{B}_x^*) \hat{z} \end{aligned}$$

But  $E_{z_{mn}} = 0$  for  $TE_{m,n}$  modes, and skipping (much) algebra:

Then:

$$\frac{1}{2\mu_o} \left( \tilde{\vec{E}}_{m,n} \times \tilde{\vec{B}}_{m,n} \right) = \frac{1}{2\mu_o} \left\{ \frac{i\pi\omega B_o^2}{(\omega/c)^2 - k_{z_{mn}}^2} \left( \frac{m}{a} \right) \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{m\pi x}{a}\right) \cos^2\left(\frac{n\pi y}{b}\right) \hat{x} \right. \\ \left. + \frac{i\pi\omega B_o^2}{(\omega/c)^2 - k_{z_{mn}}^2} \left( \frac{n}{b} \right) \cos^2\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right) \cos\left(\frac{n\pi y}{b}\right) \hat{y} \right. \\ \left. + \frac{\pi^2 \omega k_{z_{mn}} B_o^2}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left( \frac{n}{b} \right)^2 \cos^2\left(\frac{m\pi x}{a}\right) \sin^2\left(\frac{n\pi y}{b}\right) + \left( \frac{m}{a} \right)^2 \sin^2\left(\frac{m\pi x}{a}\right) \cos^2\left(\frac{n\pi y}{b}\right) \right] \hat{z} \right\}$$

Then:  $\left\langle \vec{S}_{m,n}(x, y, z, t) \right\rangle = \frac{1}{2\mu_o} \Re e \left( \tilde{\vec{E}}_{m,n}(x, y, z, t) \times \tilde{\vec{B}}_{m,n}^*(x, y, z, t) \right)$

$$\left\langle \vec{S}_{m,n}(x, y, z, t) \right\rangle = \frac{\pi^2 \omega k_{z_{mn}} B_o^2}{2\mu_o \left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left( \frac{n}{b} \right)^2 \cos^2\left(\frac{m\pi x}{a}\right) \sin^2\left(\frac{n\pi y}{b}\right) + \left( \frac{m}{a} \right)^2 \sin^2\left(\frac{m\pi x}{a}\right) \cos^2\left(\frac{n\pi y}{b}\right) \right]$$

Note that:  $\left\langle \vec{S}_{m,n}(x, y, z, t) \right\rangle = \left\langle \vec{S}_{m,n}(x, y, z, t) \right\rangle \hat{z} \leftarrow$  points in  $+\hat{z}$  direction, as it should!!

Then:  $\left\langle P_{m,n}^{trans}(z, t) \right\rangle = \int_{A_{\perp}} \left\langle \vec{S}_{m,n}(x, y, z, t) \right\rangle \cdot d\vec{a}_{\perp} = \int_{y=0}^{y=b} \int_{x=0}^{x=a} \left\langle \vec{S}_{m,n}(x, y, z, t) \right\rangle \cdot \hat{n} dx dy$   $d\vec{a}_{\perp} = \hat{n} dx dy = \hat{z} dx dy$

$$= \frac{\pi^2 \omega k_{z_{mn}} B_o^2}{2\mu_o \left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left( \frac{n}{b} \right)^2 \int_{y=0}^{y=b} \int_{x=0}^{x=a} \cos^2\left(\frac{m\pi x}{a}\right) \sin^2\left(\frac{n\pi y}{b}\right) dx dy + \left( \frac{m}{a} \right)^2 \int_{y=0}^{y=b} \int_{x=0}^{x=a} \sin^2\left(\frac{m\pi x}{a}\right) \cos^2\left(\frac{n\pi y}{b}\right) dx dy \right]$$

But:  $\int_0^a \sin^2\left(\frac{m\pi x}{a}\right) dx = \int_0^a \cos^2\left(\frac{m\pi x}{a}\right) dx = \left(\frac{a}{2}\right)$  and:  $\int_0^b \sin^2\left(\frac{n\pi y}{b}\right) dy = \int_0^b \cos^2\left(\frac{n\pi y}{b}\right) dy = \left(\frac{b}{2}\right)$

$\therefore \left\langle P_{m,n}^{trans}(z, t) \right\rangle = \frac{\pi^2 \omega k_{z_{mn}} B_o^2 ab}{8\mu_o \left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left( \frac{m}{a} \right)^2 + \left( \frac{n}{b} \right)^2 \right]$  (Watts) with:  $m = 0, 1, 2, \dots$  ( $m, n$  not both = 0)  
 $n = 0, 1, 2, \dots$  simultaneously!

But:  $k_z(\omega) = k_{z_{mn}}(\omega) = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]} = \left(\frac{\omega}{c}\right) \sqrt{1 - \left(\frac{c}{\omega}\right)^2 \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2\right]}$

Now:  $\left(\frac{\omega}{c}\right) = k_o = \frac{2\pi}{\lambda_o}$  where:  $k_o =$  vacuum wavenumber  $k_{x_m} = \left(\frac{m\pi}{a}\right)$ ,  $k_{y_n} = \left(\frac{n\pi}{b}\right)$

$\Rightarrow \lambda_o = 2\pi \left(\frac{c}{\omega}\right)$  where:  $\lambda_o =$  vacuum wavelength

Thus:

$$\left(\frac{c}{\omega}\right)^2 \left[ \left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 \right] = \left[ 2\pi \left(\frac{c}{\omega}\right) \right]^2 \left[ \left(\frac{m}{2a}\right)^2 + \left(\frac{n}{2b}\right)^2 \right] = \lambda_o^2 \left[ \left(\frac{m}{2a}\right)^2 + \left(\frac{n}{2b}\right)^2 \right] = \left[ \left(\frac{m\lambda_o}{2a}\right)^2 + \left(\frac{n\lambda_o}{2b}\right)^2 \right]$$

Thus:  $k_{z_{m,n}}(\omega) = \sqrt{k_o^2 - k_{x_m}^2 - k_{y_n}^2} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left(\frac{m\pi}{a}\right)^2 - \left(\frac{n\pi}{b}\right)^2} = k_o \sqrt{1 - \left[ \left(\frac{m\lambda_o}{2a}\right)^2 + \left(\frac{n\lambda_o}{2b}\right)^2 \right]}$

$\therefore \langle P_{m,n}^{trans}(z,t) \rangle = \frac{\omega B_o^2 ab}{8\mu_o \left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 \right] k_o \sqrt{1 - \left[ \left(\frac{m\lambda_o}{2a}\right)^2 + \left(\frac{n\lambda_o}{2b}\right)^2 \right]}$

or:  $\langle P_{m,n}^{trans}(z,t) \rangle = \frac{1}{2\mu_o \omega} \frac{\omega^2 B_o^2 (\frac{1}{2}a)(\frac{1}{2}b)}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 \right] k_o \sqrt{1 - \left[ \left(\frac{m\lambda_o}{2a}\right)^2 + \left(\frac{n\lambda_o}{2b}\right)^2 \right]}$

But:  $\int_{A_1} \left\langle \left| \tilde{E}_x(x,y,z,t) \right|^2 \right\rangle dx dy = \frac{\omega^2 k_{y_n}^2 B_o^2 (\frac{1}{2}a)(\frac{1}{2}b)}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} = \frac{\omega^2 B_o^2 (\frac{1}{2}a)(\frac{1}{2}b)}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \left(\frac{n\pi}{b}\right)^2$  with:  $k_{y_n} \equiv \left(\frac{n\pi}{b}\right)$

$\int_{A_1} \left\langle \left| \tilde{E}_y(x,y,z,t) \right|^2 \right\rangle dx dy = \frac{\omega^2 k_{x_m}^2 B_o^2 (\frac{1}{2}a)(\frac{1}{2}b)}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} = \frac{\omega^2 B_o^2 (\frac{1}{2}a)(\frac{1}{2}b)}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \left(\frac{m\pi}{a}\right)^2$  with:  $k_{x_m} \equiv \left(\frac{m\pi}{a}\right)$

Defining:

$$\left. \begin{aligned} |\tilde{E}_{o_x}^{m,n}| &\equiv \frac{\omega k_{y_n} B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} = \frac{\omega B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \left(\frac{n\pi}{b}\right) \\ |\tilde{E}_{o_y}^{m,n}| &\equiv \frac{\omega k_{x_m} B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} = \frac{\omega B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \left(\frac{m\pi}{a}\right) \end{aligned} \right\} \begin{array}{l} \text{Magnitudes} \\ \text{of } \hat{x}, \hat{y} \\ \text{electric field} \\ \text{amplitudes} \end{array}$$

Then:  $\langle P_{m,n}^{trans}(z,t) \rangle = \frac{k_o}{2\mu_o \omega} \left( |\tilde{E}_{o_x}^{m,n}|^2 + |\tilde{E}_{o_y}^{m,n}|^2 \right) \left(\frac{a}{2}\right) \left(\frac{b}{2}\right) \sqrt{1 - \left[ \left(\frac{m\lambda_o}{2a}\right)^2 + \left(\frac{n\lambda_o}{2b}\right)^2 \right]}$

But:  $\frac{k_o}{\omega} = \frac{1}{c} \Rightarrow \frac{k_o}{\mu_o \omega} = \frac{1}{\mu_o} \left(\frac{k_o}{\omega}\right) = \frac{1}{\mu_o c}$  but:  $c = \sqrt{\frac{1}{\epsilon_o \mu_o}} \therefore \frac{k_o}{\mu_o \omega} = \frac{1}{\mu_o c} = \frac{\sqrt{\epsilon_o \mu_o}}{\mu_o} = \sqrt{\frac{\epsilon_o}{\mu_o}}$

$\therefore \langle P_{m,n}^{trans}(z,t) \rangle = \frac{1}{2} \sqrt{\frac{\epsilon_o}{\mu_o}} \left( \frac{1}{4} |\tilde{E}_{o_x}^{m,n}|^2 + \frac{1}{4} |\tilde{E}_{o_y}^{m,n}|^2 \right) ab \sqrt{1 - \left[ \left(\frac{m\lambda_o}{a}\right)^2 + \left(\frac{n\lambda_o}{b}\right)^2 \right]}$   $\lambda_o \equiv$  vacuum wave length  $= c/f$

$\Rightarrow$  The time-averaged power transported down the hollow rectangular waveguide for the  $TE_{mn}^{th}$  mode is proportional to the square of the  $E$ -field amplitudes in the  $\hat{x}$  and  $\hat{y}$  direction!!

We introduce the parameter  $Z_o(\vec{r}) \equiv \left| \frac{\tilde{E}_\perp(\vec{r})}{\tilde{B}_\perp(\vec{r})/\mu_o} \right| = \mu_o c = \sqrt{\mu_o/\epsilon_o} = 120\pi \Omega \approx 377\Omega$

= the *EM wave impedance of free space* (*n.b.*  $Z_o$  is a purely real, scalar quantity because there is no dissipation!). For  $TE_{mn}$  modes of *EM wave propagation in a waveguide that has perfectly conducting walls* (*i.e.* no dissipation/no losses); the *EM wave impedance of the waveguide* is also purely real:

$$Z_{TE}^{m,n}(\omega) \equiv \left| \frac{\tilde{E}_{TE}^\perp(\vec{r})}{\tilde{B}_{TE}^\perp(\vec{r})/\mu_o} \right| = Z_o(\lambda_z^{m,n}(\omega)/\lambda_o) = Z_o(k_o/k_{z_{m,n}}(\omega)).$$

Then since:

$$k_{z_{m,n}}(\omega) = \frac{2\pi}{\lambda_z^{m,n}(\omega)} = \sqrt{\left(\frac{\omega}{c}\right)^2 - \left(\frac{m\pi}{a}\right)^2 - \left(\frac{n\pi}{b}\right)^2} = k_o \sqrt{1 - \left[\left(\frac{m\lambda_o}{2a}\right)^2 + \left(\frac{n\lambda_o}{2b}\right)^2\right]} = \sqrt{k_o^2 - k_{x_m}^2 - k_{y_n}^2} < k_o$$

or equivalently  $\lambda_z^{m,n}(\omega) > \lambda_o$  we see that:  $Z_{TE}^{m,n}(\omega) = Z_o(\lambda_z^{m,n}(\omega)/\lambda_o) = Z_o(k_o/k_{z_{m,n}}(\omega)) > 377\Omega$  for  $TE_{m,n}$  modes of *EM wave propagation in a waveguide*.

We can thus write the *EM power transmitted down the waveguide* for  $TE_{mn}$  modes as:

$$\langle P_{m,n}^{trans}(z,t) \rangle = \frac{1}{2Z_o} \left( \frac{1}{4} |\tilde{E}_{o_x}^{m,n}|^2 + \frac{1}{4} |\tilde{E}_{o_y}^{m,n}|^2 \right) A_\perp \left( \frac{\lambda_o}{\lambda_z^{m,n}(\omega)} \right) = \frac{1}{2} \left( \frac{1}{4} |\tilde{E}_{o_x}^{m,n}|^2 + \frac{1}{4} |\tilde{E}_{o_y}^{m,n}|^2 \right) A_\perp / Z_{TE}^{m,n}(\omega)$$

Where  $A_\perp = ab$  = cross-sectional area of the rectangular waveguide.

Note that this expression is analogous to  $\langle P \rangle = \frac{1}{2} V_{peak}^2 / R$ , since  $E^2 A_\perp \sim (\text{Volts/m})^2 * m^2 = \text{Volts}^2$ .

## The Energy Density $\langle u_{m,n} \rangle$ Stored in a Rectangular Waveguide - TE<sub>m,n</sub> Mode

Again, from Griffiths Problem 9.11 (p. 382) since  $\langle fg \rangle = \frac{1}{2} \text{Re}(\tilde{f} \tilde{g}^*)$

Then: 
$$\langle u \rangle = \frac{1}{4} \text{Re} \left\{ \epsilon_o \tilde{\mathbf{E}} \cdot \tilde{\mathbf{E}}^* + \frac{1}{\mu_o} \tilde{\mathbf{B}} \cdot \tilde{\mathbf{B}}^* \right\}$$

Then in the ( $m,n$ <sup>th</sup>) TE mode: 
$$\langle u_{m,n} \rangle = \frac{1}{4} \text{Re} \left\{ \epsilon_o \tilde{\mathbf{E}}_{m,n} \cdot \tilde{\mathbf{E}}_{m,n}^* + \frac{1}{\mu_o} \tilde{\mathbf{B}}_{m,n} \cdot \tilde{\mathbf{B}}_{m,n}^* \right\}$$

Where: 
$$\tilde{\mathbf{E}}_{m,n} = \tilde{E}_{x_{m,n}} \hat{x} + \tilde{E}_{y_{m,n}} \hat{y} + \tilde{E}_{z_{m,n}} \hat{z} \quad \text{0 for TE modes}$$

And: 
$$\tilde{\mathbf{B}}_{m,n} = \tilde{B}_{x_{m,n}} \hat{x} + \tilde{B}_{y_{m,n}} \hat{y} + \tilde{B}_{z_{m,n}} \hat{z} \quad \text{0 for TE modes}$$

n.b.  $|A|^2 \equiv A \cdot A^*$

Then: 
$$\langle u_{m,n} \rangle = \frac{1}{4} \text{Re} \left\{ \epsilon_o \left[ |E_{x_{m,n}}|^2 + |E_{y_{m,n}}|^2 + |E_{z_{m,n}}|^2 \right] + \frac{1}{\mu_o} \left[ |B_{x_{m,n}}|^2 + |B_{y_{m,n}}|^2 + |B_{z_{m,n}}|^2 \right] \right\}$$

$$\begin{aligned} \langle u_{m,n} \rangle = & \frac{\epsilon_o}{4} \left( \frac{\omega \pi B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \right)^2 \left[ \left( \frac{n}{b} \right)^2 \cos^2 \left( \frac{m\pi x}{a} \right) \sin^2 \left( \frac{n\pi y}{b} \right) + \left( \frac{m}{a} \right)^2 \sin^2 \left( \frac{m\pi x}{a} \right) \cos^2 \left( \frac{n\pi y}{b} \right) \right] \\ & + \frac{1}{4\mu_o} \left\{ \left( \frac{k_{m,n} \pi B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \right)^2 \left[ \left( \frac{n}{b} \right)^2 \cos^2 \left( \frac{m\pi x}{a} \right) \sin^2 \left( \frac{n\pi y}{b} \right) + \left( \frac{m}{a} \right)^2 \sin^2 \left( \frac{m\pi x}{a} \right) \cos^2 \left( \frac{n\pi y}{b} \right) \right] \right. \\ & \left. + B_o^2 \cos^2 \left( \frac{m\pi x}{a} \right) \cos^2 \left( \frac{n\pi y}{b} \right) \right\} \end{aligned}$$

The time-averaged energy per unit length (Joules/m) in the waveguide for the ( $m,n$ <sup>th</sup>) TE mode is:

$$\langle E_{m,n} \rangle / L \equiv \int_{A_\perp} \langle u_{m,n} \rangle da_\perp = \int_{y=0}^{y=b} \int_{x=0}^{x=a} \langle u_{m,n} \rangle dx dy$$
 where  $L$  (meters) is the length of the waveguide.

$$\begin{aligned} \langle E_{m,n} \rangle / L \equiv \int_{A_\perp} \langle u_{m,n} \rangle da_\perp = & \frac{\epsilon_o}{4} \left( \frac{\pi \omega B_o}{\left[ (\omega/c)^2 - k_{m,n}^2 \right]} \right)^2 \left( \frac{a}{2} \right) \left( \frac{b}{2} \right) \left[ \left( \frac{n}{b} \right)^2 + \left( \frac{m}{a} \right)^2 \right] \\ & + \frac{1}{4\mu_o} \left\{ \left( \frac{\pi k_{m,n} B_o}{\left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]} \right)^2 \left( \frac{a}{2} \right) \left( \frac{b}{2} \right) \left[ \left( \frac{n}{b} \right)^2 + \left( \frac{m}{a} \right)^2 \right] + \left( \frac{a}{2} \right) \left( \frac{b}{2} \right) B_o^2 \right\} \end{aligned}$$

$$\langle E_{m,n} \rangle / L \equiv \int_{A_{\perp}} \langle u_{m,n} \rangle da_{\perp} = \frac{\epsilon_0}{4} \left( \frac{\pi \omega B_o}{(\omega/c)^2 - k_{z_{mn}}^2} \right)^2 \left( \frac{ab}{4} \right) \left[ \left( \frac{n}{b} \right)^2 + \left( \frac{m}{a} \right)^2 \right] + \frac{1}{4\mu_0} \left\{ \left( \frac{\pi k_{m,n} B_o}{(\omega/c)^2 - k_{z_{mn}}^2} \right)^2 \left( \frac{ab}{4} \right) \left[ \left( \frac{n}{b} \right)^2 + \left( \frac{m}{a} \right)^2 \right] + \left( \frac{ab}{4} \right) B_o^2 \right\}$$

Now: 
$$\langle P_{m,n}^{trans}(z,t) \rangle = \int_{A_{\perp}} \langle \vec{S}_{m,n} \rangle \cdot d\vec{a}_{\perp} = \frac{\omega k_{m,n} B_o^2 ab}{8\mu_0 \left[ (\omega/c)^2 - k_{z_{mn}}^2 \right]^2} \left[ \left( \frac{m\pi}{a} \right)^2 + \left( \frac{n\pi}{b} \right)^2 \right]$$

and: 
$$k_{z_{mn}}^2 = \left( \frac{\omega}{c} \right)^2 - \left[ \left( \frac{m\pi}{a} \right)^2 + \left( \frac{n\pi}{b} \right)^2 \right] \Rightarrow \left[ \left( \frac{\omega}{c} \right)^2 - k_{z_{mn}}^2 \right] = \left[ \left( \frac{m\pi}{a} \right)^2 + \left( \frac{n\pi}{b} \right)^2 \right] \equiv \left( \frac{\omega_{m,n}^2}{c^2} \right)$$

Thus: 
$$\langle P_{m,n}^{trans} \rangle = \int_{A_{\perp}} \langle \vec{S}_{m,n} \rangle \cdot d\vec{a}_{\perp} = \frac{\omega k_{z_{mn}} ab}{8\mu_0 \omega_{mn}^2} c^2 B_o^2$$

And: 
$$\frac{\langle E_{m,n} \rangle}{L} = \int_{A_{\perp}} \langle u_{m,n} \rangle da_{\perp} = \frac{\omega^2 ab}{8\mu_0 \omega_{mn}^2} B_o^2$$

Note that the ratio of: 
$$\frac{\langle P_{m,n}^{trans} \rangle}{\langle E_{m,n} \rangle / L} = \frac{\text{Watts}}{\text{Joules/m}} = \frac{\text{Joules/sec}}{\text{Joules/m}} = \frac{m}{\text{sec}} \quad (\text{i.e. speed})$$

$$\frac{\langle P_{m,n}^{trans} \rangle}{\langle E_{m,n} \rangle / L} = \frac{\frac{\omega k_{z_{mn}} ab}{8\mu_0 \omega_{mn}^2} c^2 B_o^2}{\frac{\omega^2 ab}{8\mu_0 \omega_{mn}^2} B_o^2} = \frac{k_{m,n} c^2}{\omega} = \left( \frac{c}{\omega} \right) (k_{m,n} c)$$

But: 
$$k_{z_{mn}}^2 c^2 = \omega^2 - \omega_{mn}^2 \quad \text{or:} \quad k_{z_{mn}} c = \sqrt{\omega^2 - \omega_{mn}^2}$$

$$\therefore \frac{\langle P_{m,n}^{trans} \rangle}{\langle E_{m,n} \rangle / L} = \left( \frac{c}{\omega} \right) \sqrt{\omega^2 - \omega_{mn}^2} = c \sqrt{1 - \left( \frac{\omega_{mn}}{\omega} \right)^2} = v_{g_z}^{mn}(\omega) \quad !!!$$

or: 
$$v_{g_z}^{mn}(\omega) = c \sqrt{1 - \left( \frac{\omega_{mn}}{\omega} \right)^2} = \frac{\langle P_{m,n}^{trans} \rangle}{\langle E_{m,n} \rangle / L} \quad \text{Thus we see that:} \quad \langle P_{m,n}^{trans} \rangle = v_{g_z}^{m,n}(\omega) \cdot \langle E_{m,n} \rangle / L$$

## Propagation of TM Waves in a Hollow Rectangular Waveguide

For propagation TM waves in a hollow waveguide,  $\tilde{E}_z \neq 0$ , but  $\tilde{B}_z = 0$ .

$\therefore$  We need to solve 3-D wave equation: 
$$\left[ \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \left( \frac{\omega}{c} \right)^2 - k_z^2 \right] E_z = 0$$

subject to boundary conditions on the inner walls of rectangular waveguide:  $\tilde{E}_{\parallel} = 0$  and  $\tilde{B}_{\perp} = 0$

Following the procedure we developed for the TE mode case, let:  $\tilde{E}_z(x, y) = \tilde{X}(x)\tilde{Y}(y)$

$$\begin{aligned} \tilde{X}(x) = \tilde{A}_x \cos k_x x + \tilde{B}_x \sin k_x x = 0 \text{ \{at } x=0 \text{ and } x=a\} &\Rightarrow \tilde{A}_x = 0 &\Rightarrow \tilde{X}(x) = \tilde{B}_x \sin k_x x \\ \tilde{Y}(y) = \tilde{A}_y \cos k_y y + \tilde{B}_y \sin k_y y = 0 \text{ \{at } y=0 \text{ and } y=b\} &\Rightarrow \tilde{A}_y = 0 &\Rightarrow \tilde{Y}(y) = \tilde{B}_y \sin k_y y \end{aligned}$$

Because  $m, n = 0, 1, 2, 3, \dots$  the lowest non-trivial  $TM_{mn}$  mode is  $TM_{11}$

$$\left\{ \begin{aligned} k_x &\equiv \left( \frac{m\pi}{a} \right), \quad m = 1, 2, 3, \dots && \text{n.b. } m = 0 \text{ is NOT allowed here!!! } (\Rightarrow X(x) \equiv 0 \text{ everywhere!!!}) \\ k_y &\equiv \left( \frac{n\pi}{b} \right), \quad n = 1, 2, 3, \dots && \text{n.b. } n = 0 \text{ is NOT allowed here!!! } (\Rightarrow Y(y) \equiv 0 \text{ everywhere!!!}) \end{aligned} \right.$$

Then:  $\tilde{E}_z(x, y) = \tilde{E}_o \sin(k_x x) \sin(k_y y) = \tilde{E}_o \sin\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right)$  with  $m, n = 1, 2, 3, \dots$

All the rest is the {nearly} the same as for TE waves –  $\langle P_{m,n}^{trans} \rangle, \langle u_{m,n} \rangle, \text{ etc.}$

Then:  $k_{z_{mn}} = \sqrt{(\omega/c)^2 - (m\pi/a)^2 - (n\pi/b)^2}$  (as before)

The cutoff angular frequency:  $\omega_{mn} \equiv c \sqrt{(m\pi/a)^2 + (n\pi/b)^2} \Rightarrow k_{z_{mn}} = \frac{1}{c} \sqrt{\omega^2 - \omega_{mn}^2}$

Phase (aka wave) velocity:  $v_{\phi_z}^{m,n} = c \frac{1}{\sqrt{1 - (\omega_{m,n}/\omega)^2}}$  Group velocity:  $v_{g_z}^{m,n}(\omega) = c \sqrt{1 - (\omega_{m,n}/\omega)^2}$

One difference for TM modes vs. TE modes is that the wave impedance is:

$$Z_{TM}^{m,n}(\omega) = Z_o \left( \lambda_o / \lambda_z^{m,n}(\omega) \right) = Z_o \left( k_{z_{mn}}(\omega) / k_o \right) \quad \text{vs.} \quad Z_{TE}^{m,n}(\omega) = Z_o \left( \lambda_z^{m,n}(\omega) / \lambda_o \right) = Z_o \left( k_o / k_{z_{mn}}(\omega) \right).$$

Since  $\lambda_z^{mn}(\omega) > \lambda_o$  and  $Z_o \equiv \sqrt{\mu_o / \epsilon_o} = 120\pi \Omega \approx 377\Omega$  then we see that:

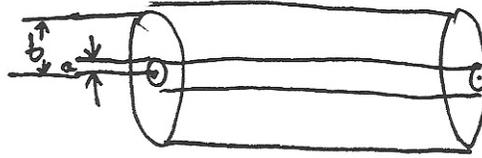
$$Z_{TM}^{m,n}(\omega) = Z_o \left( \lambda_o / \lambda_z^{m,n}(\omega) \right) < 377\Omega \quad \text{whereas:} \quad Z_{TE}^{m,n}(\omega) = Z_o \left( \lambda_z^{m,n}(\omega) / \lambda_o \right) > 377\Omega$$

The ratio of the lowest TM mode to the lowest TE mode is:

$$\left( \omega_{m,n}^{TM} / \omega_{m,n}^{TE} \right) = \left( \omega_{11}^{TM} / \omega_{10}^{TE} \right) = \sqrt{(1/a)^2 + (1/b)^2} / \sqrt{(1/a)^2} = \sqrt{1 + (a/b)^2}$$

## Propagation of TEM Waves in a Coaxial Transmission Line

- We have previously shown that a hollow waveguide **cannot** support TEM waves  $\{\tilde{E}_z = \tilde{B}_z = 0\}$
- However, a coaxial transmission line, consisting of an inner, long straight wire of radius  $a$ , surrounded by a cylindrical conducting sheath of radius  $b > a$  **does** support the propagation of TEM waves:



For TEM waves:  $k = \omega/c$ . TEM waves travel at speed of light  $c \Rightarrow$  non-dispersive!

For TEM waves, Maxwell's equations give:

(1) Gauss' Law:  $\vec{\nabla} \cdot \tilde{\mathbf{E}} = 0$

$$\frac{\partial \tilde{E}_{ox}}{\partial x} + \frac{\partial \tilde{E}_{oy}}{\partial y} = 0$$

$$\Rightarrow \boxed{\frac{\partial \tilde{E}_{ox}}{\partial x} = -\frac{\partial \tilde{E}_{oy}}{\partial y}}$$

(2) No monopoles:  $\vec{\nabla} \cdot \tilde{\mathbf{B}} = 0$

$$\frac{\partial \tilde{B}_{ox}}{\partial x} + \frac{\partial \tilde{B}_{oy}}{\partial y} = 0$$

$$\Rightarrow \boxed{\frac{\partial \tilde{B}_{ox}}{\partial x} = -\frac{\partial \tilde{B}_{oy}}{\partial y}}$$

(3) Faraday's Law:  $\left( \vec{\nabla} \times \tilde{\mathbf{E}} = -\frac{\partial \tilde{\mathbf{B}}}{\partial t} \right)$

(i)  $\frac{\partial \tilde{E}_{oy}}{\partial x} - \frac{\partial \tilde{E}_{ox}}{\partial y} = i\omega \tilde{B}_{oz} = 0$

(ii)  $\frac{\partial \tilde{E}_{oz}}{\partial y} - ik\tilde{E}_{oy} = i\omega \tilde{B}_{ox} = 0$

(iii)  $ik\tilde{E}_{ox} - \frac{\partial \tilde{E}_{oz}}{\partial x} = i\omega \tilde{B}_{oy} = 0$

(4) Ampere's Law:  $\left( \vec{\nabla} \times \tilde{\mathbf{B}} = \frac{1}{c^2} \frac{\partial \tilde{\mathbf{E}}}{\partial t} \right)$

(iv)  $\frac{\partial \tilde{B}_{oy}}{\partial x} - \frac{\partial \tilde{B}_{ox}}{\partial y} = -\frac{i\omega}{c^2} \tilde{E}_{oz} = 0$

(v)  $\frac{\partial \tilde{B}_{oz}}{\partial y} = -\frac{i\omega}{c^2} \tilde{E}_{ox}$

(vi)  $ik\tilde{B}_{ox} - \frac{\partial \tilde{B}_{oz}}{\partial x} = -\frac{i\omega}{c^2} \tilde{E}_{oy}$

which can be rewritten:

(i)  $\boxed{\frac{\partial \tilde{E}_{oy}}{\partial x} = \frac{\partial \tilde{E}_{ox}}{\partial y}}$

(ii)  $\boxed{\tilde{B}_{ox} = -\frac{k}{\omega} \tilde{E}_{oy} = -\frac{1}{c} \tilde{E}_{oy}}$

(iii)  $\boxed{\tilde{B}_{oy} = \frac{k}{\omega} \tilde{E}_{ox} = +\frac{1}{c} \tilde{E}_{ox}}$

(iv)  $\boxed{\frac{\partial \tilde{B}_{oy}}{\partial x} = \frac{\partial \tilde{B}_{ox}}{\partial y}}$

(v)  $\boxed{\tilde{B}_{oy} = \frac{\omega}{c^2 k} \tilde{E}_{ox} = \frac{1}{c} \tilde{E}_{ox}}$

(vi)  $\boxed{\tilde{B}_{ox} = -\frac{\omega}{c^2 k} \tilde{E}_{oy} = -\frac{1}{c} \tilde{E}_{oy}}$

Note that equations (iii) and (v) above give the same relation  $\tilde{B}_{oy} = \frac{1}{c} \tilde{E}_{ox}$

as do equations (ii) and (vi),  $\tilde{B}_{ox} = -\frac{1}{c} \tilde{E}_{oy}$ .

The following six relations:

$\tilde{B}_{oy} = \frac{1}{c} \tilde{E}_{ox}$	$\tilde{B}_{ox} = -\frac{1}{c} \tilde{E}_{oy}$
$\frac{\partial \tilde{E}_{ox}}{\partial y} = \frac{\partial \tilde{E}_{oy}}{\partial x}$	$\frac{\partial \tilde{B}_{ox}}{\partial y} = \frac{\partial \tilde{B}_{oy}}{\partial x}$
$\frac{\partial \tilde{E}_{ox}}{\partial x} = -\frac{\partial \tilde{E}_{oy}}{\partial y}$	$\frac{\partial \tilde{B}_{ox}}{\partial x} = -\frac{\partial \tilde{B}_{oy}}{\partial y}$

Are precisely the same equations of electrostatics and magnetostatics for empty space (*i.e.* the vacuum) in two dimensions.

Since a coaxial cable has cylindrical geometry/cylindrical symmetry, the TEM electric field (as in case of the infinite line charge) must be of the form:

$$\tilde{\tilde{E}}_o(\rho, \varphi) = \frac{\tilde{A}}{\rho} \hat{\rho} \quad \text{where } \tilde{A} = \text{constant.}$$

Similarly, the TEM magnetic field (as in the case of infinite line current) must be of the form:

$$\tilde{\tilde{B}}_o(\rho, \varphi) = \frac{\tilde{A}}{\rho c} \hat{\phi}$$

Then for TEM wave propagation in a coaxial transmission line:

$\tilde{\tilde{E}}(\rho, \varphi, z, t) = \tilde{\tilde{E}}_o(\rho, \varphi) e^{i(kz - \omega t)} = \frac{\tilde{A}}{\rho} e^{i(kz - \omega t)} \hat{\rho}$	$\Rightarrow$ <b><u>no</u></b> dispersion!
$\tilde{\tilde{B}}(\rho, \varphi, z, t) = \tilde{\tilde{B}}_o(\rho, \varphi) e^{i(kz - \omega t)} = \frac{\tilde{A}}{\rho c} e^{i(kz - \omega t)} \hat{\phi}$	
$k = \frac{\omega}{c}$	$v_{group} = v_{phase} = c$

Note that there are **no** restrictions on the value of  $k$  for TEM waves in a coaxial cable.

For TEM *EM* wave propagation in a coaxial transmission line that has perfectly conducting walls (*i.e.* no dissipation/no losses), the *EM* wave impedance is (again) purely real:

$$Z_{TEM}^{coax}(\omega) \equiv \left| \frac{\tilde{\tilde{E}}_{TEM}^\perp(\vec{r})}{\tilde{\tilde{B}}_{TEM}^\perp(\vec{r}) / \mu_o} \right| = \sqrt{\mu_o / \epsilon_o} = Z_o = 120\pi \Omega = 377\Omega.$$